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Donut-Shaped mmWave Printed Antenna Array for 5G Technology

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Abstract: This article presents compact and novel shape ring-slotted antenna array operating at mmwave band on central frequency of 28 GHz. The proposed structure designed at 0.256 mm thin Rogers 5880 is composed of a ring shape patch with a square slot etched at the top mid-section of partial ground plane. Through optimizing the ring and square slot parameters, a high bandwidth of 8 GHz is achieved, ranging from 26 to 32 GHz, with a simulated gain of 3.95 dBi and total efficiency of 96% for a single element. The proposed structure is further transformed in a 4-element linear array manner. With compact dimensions of 20 mm × 22 mm for array, the proposed antenna delivers a high simulated gain of 10.7 dBi and is designed in such a way that it exhibits dual beam response over the entire band of interest and simulated results agree with fabricated prototype measurements.

Keywords: 28 GHz; mmwave band; directivity; gain; 5G; smart phones

1. Introduction

5G technology offers immensely high data rate characteristics with low latency over a whole communication channel. Divided into two bands, 5G technology focuses on sub 6 GHz with different sub bands, such as LTE band 42 (3.4–3.6), band 43 (2.6–2.8), etc., but Sub6GHz spectrum is already choking up with an enormous burden of microwave applications. This leads to the second dedicated mmWave spectrum which has unlimited bandwidth to be utilized [1,2]. With available feasibility reports, it is noteworthy that mmWave range communication is a promising candidate for providing wider bandwidth, low latency, and higher data throughputs. In mmWave band, numerous bands have been assigned as a possible candidate for future 5G standards, such as, 28 GHz and 38 GHz band, while 57–64 GHz (O2 band), and 164–200 (H2O band) have been declared as unlicensed frequency bands [3,4].

An infinity shell shape four element multiple input multiple output (MIMO) antenna array is presented in [5]. This system is operating at the central frequency of 28 GHz and is designed for mmWave 5G applications. In this work, the elements are arranged in such a

manner that they provide spatial and pattern diversity characteristics, but with low realized gain of 5.5 dBi. Similarly, in [6], a mmWave antenna array is presented, featuring a dual band on 28 and 38 GHz band. The four-element MIMO array is arranged in an orthogonal manner providing directive patterns at both dual resonances, but the bandwidth response is extremely low for 5G applications and services. In mmWave band, the attenuation and path losses become critical, since the wavelength becomes extremely small. These issues can be overcome with high-gain antenna arrays compensating for propagation and attenuation losses. Substrate integrated waveguides (SIW) are known to have low transmission line losses. A Hexagon loop shape SIW four element array is presented in [7], which provides broadband radiation characteristics with a high gain of 12 dBi. Similarly, in [8], a large impedance bandwidth of 16.5% at 28 GHz antenna is reported with SIW cavity-backed configuration. The overall size of array is small ($23 \text{ mm} \times 22.4 \text{ mm}$) but the use of combination of series and parallel SIW-feed mechanism to excite different modes has made the reported structure complex by nature. In [9], an 8-element SIW cavity backed antenna array is presented with extremely low side-lobe levels. The reported antenna provides good radiation patterns and gain of around 13 dBi but the bandwidth of 2.3 GHz is attained only with a large size of $63.5 \times 70 \text{ mm}^2$.

Planar antenna arrays, on the other hand, provide simple structure characteristics and are easy to assemble which of these have transmission line losses, but they can be neglected with careful modelling [10–14]. A 9-element phased array is reported in [15], which exhibits a gain of nearly 15 dBi. The structure is simple with a small size of 400 mm^2 , but the offered bandwidth is very low. In a similar study [16], a simple 3×3 planar microstrip-phased array antenna operating at 28 GHz is presented. It has a high gain of 15.6 dBi and 1.7 GHz bandwidth, with a small $20 \times 20 \text{ mm}^2$ size. A two-element MIMO planar structure in [17] with relative size of $26 \times 11 \text{ mm}^2$ resonates at two different mmWave allocated bands of 27 GHz and 39 GHz, but the offered gain of the antenna is 5 dB, which is low for mmWave systems. For mmWave systems, wide bandwidth and high gain are necessary in order to overcome atmospheric pressures and multipath fading effects.

As discussed earlier, for 5G technology characteristics, such as, high gain, wide bandwidth, and low latency, it is necessary to overcome challenges, such as, multipath fading and atmospheric pressure, therefore it is reasonable to study and propose a system that can address these challenges efficiently, and this is the motivation behind this work. In this work, a low-cost, simple, and easy to fabricate, non-complex, and integrateable planar structure with simple microstrip feed network is proposed. This research presents a novel, yet simple planar structure with simple feed network of 4 element array for mmWave applications. The emphasis here is on a simple and low-cost system that should not add complexity to the system when integrated. In literature review, we have discussed that although SIW based system are available but they are complex and hard to integrate. Therefore, there is need of such simple and easy to fabricate antenna systems. It is worthy to mention that, this system has dual beam radiation response on the front and the back face, and a peak gain of above 11 dBi, therefore it is believed that the proposed array can find its applications in the future mmWave broadband applications.

2. Antenna Design

Before presenting the proposed antenna array as shown in Figure 1, first, design, working principle, and results for a single element are presented. The aim here is to understand the working of the proposed system at the fundamental level and discuss how the final product is achieved. The single element is designed on a Rogers-5880 substrate having a relative permittivity of 2.2, and loss tangent of 0.0009. The total size of the single element is $8 \times 10 \times 0.254 \text{ mm}^3$, while the thickness of the copper is 0.035 mm. The single element consists of a feed line, a donut-shaped radiating element, and a partial square-slotted ground plane. The design parameters for the single element are given in Table 1.

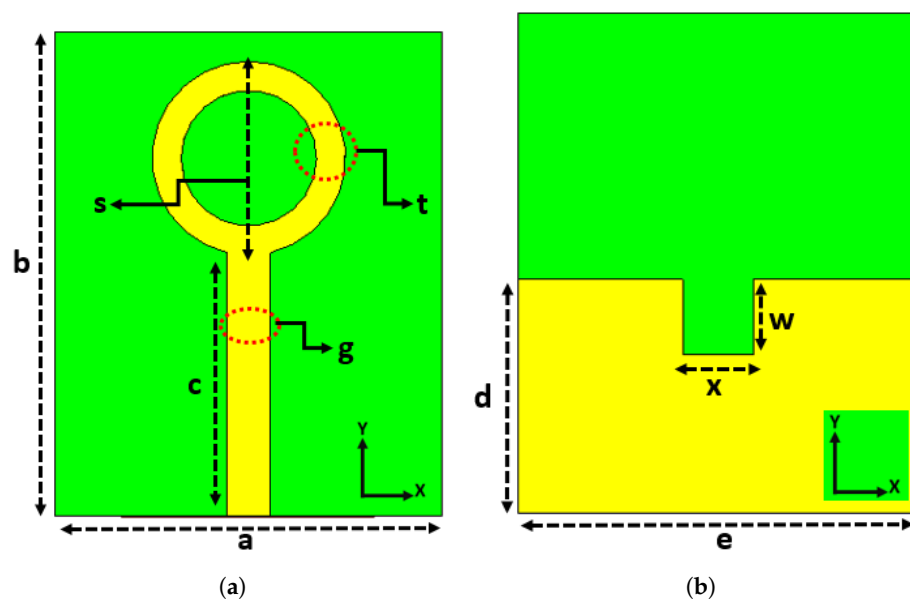


Figure 1. Proposed Antenna Design. (a) Front View. (b) Back View.

Table 1. Design Parameters.

Parameter	Value in mm	Parameter	Value in mm
a	8	b	10
c	6	g	0.9
s	1.5	t	0.6
d	4.65	e	8
w	1.5	x	1.5

Figure 2 illustrates the evolution of the single elements. First, a stub is designed and reflection coefficient is observed. Then, a circle is added at the end of the stub to make the antenna resonate within the desired frequency range. Finally, another circle is added within the previous circle and subtracted. This design resonates at 28 GHz with 6 GHz bandwidth ranges from 26 GHz to 32 GHz, as shown in Figure 2. Please note that, while designing the model, the ground plane was partial and a square-shaped slot was removed. This single element works as a simple circular patch antenna but the ground plane is partial. This is done to achieve dual radiation patterns within a desired frequency range.

The surface current distribution for a single element is shown in Figure 3. Next, a few parametric studies are performed to learn the effects of different design variables on the performance of the antenna specifically, on the S-parameters, as shown in Figure 4. First, the width between the two circles is varied to see the change in the scattering parameters. The variable t is varied between 1.2 mm and 1.4 mm with an interval of 0.1 mm. It is observed that as we increase the width, the frequency shifts from higher end of the band to the lower of the band and also the impedance of the antenna improves providing better matching, as depicted in Figure 4a. Similarly, to understand the effect of the rectangular shaped slot in the ground plane, a variable x is varied for five different values ranges between 1.3 mm to 1.7 mm.

It is found that this variable can be used to obtain better impedance matching between the antenna and the feeding port, as illustrated in Figure 4b. Finally, the length of the feed is studied and its effect on the performance of the antenna is depicted in Figure 4c. As we increase the length of the feed, the antenna resonates at the lower side of the band. This is due to the fact that the electrical length and the physical length of the antenna is being increase and that is why it is resonating at the lower frequency. Once the parametric study

is done and optimum parameters are selected, then the s-parameters of the final single element are shown in Figure 4d.

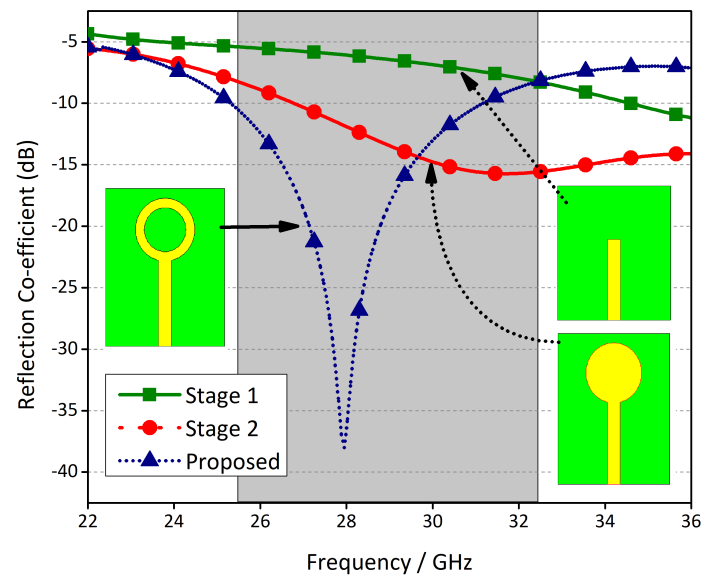


Figure 2. Four stage Design Evolution.

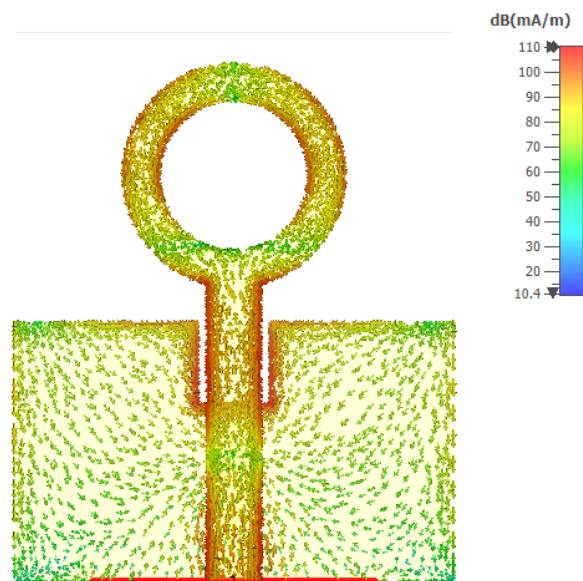


Figure 3. Surface Current Distributions.

The radiation and total efficiency and gain of the proposed single element are shown in Figure 5. Please note that the efficiency is more than 90%, while the gain is around 4 dBi. Next, different key performance parameters of the proposed antenna array are discussed in detail.

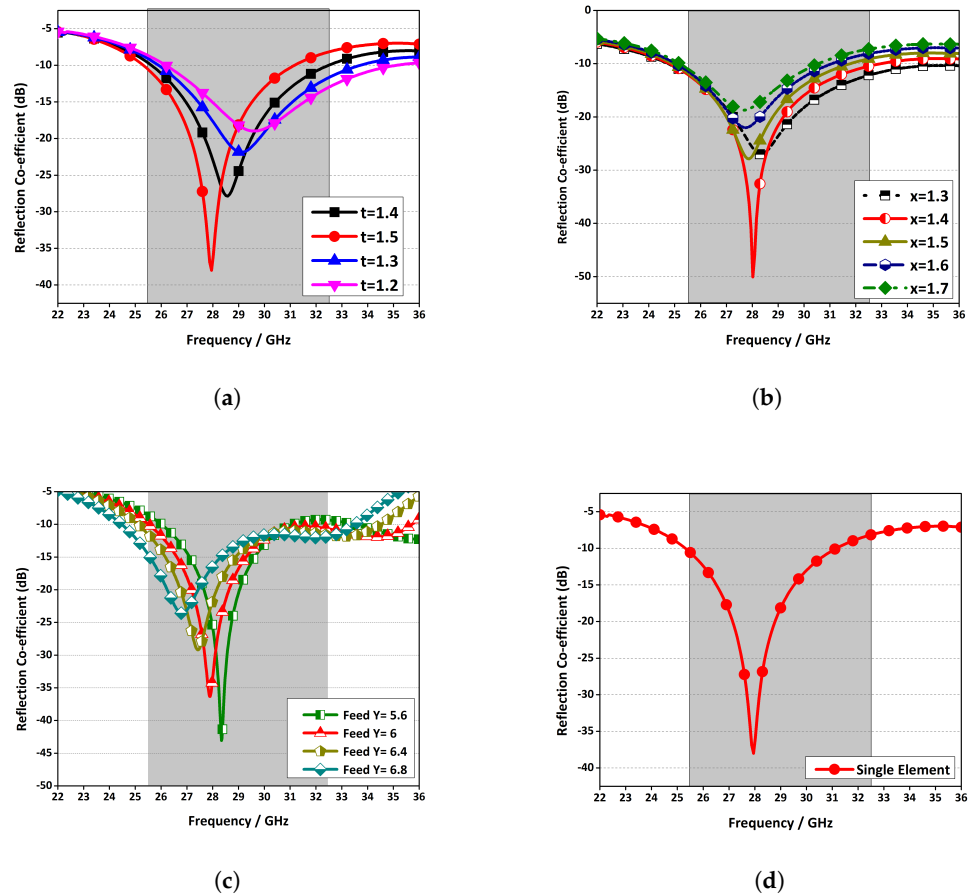


Figure 4. Reflection Co-efficient. (a) Strip 1 sweep. (b) Strip 2 sweep. (c) Feed length sweep. (d) Final Single Element.

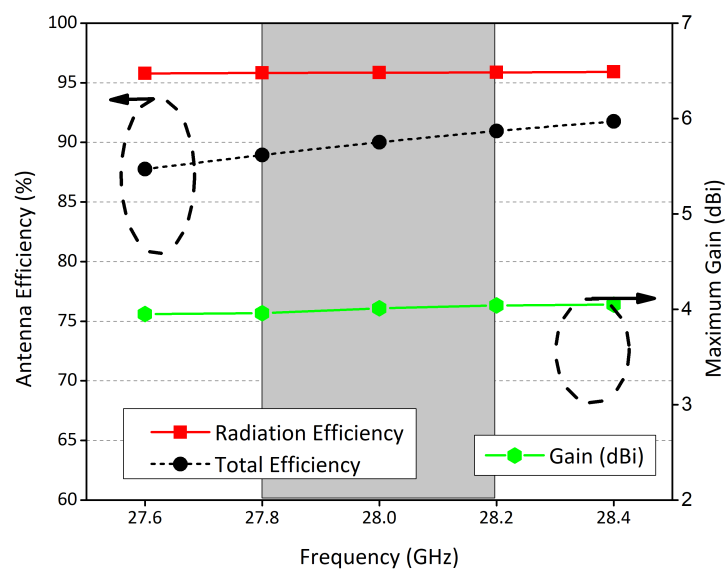


Figure 5. Performance parameters of the proposed Antenna.

As discussed earlier, based on the reasonable performance of the single element, a linear array is designed using a simple feed structure. The feed line consists of a $50\ \Omega$ transmission line that divides into parallel connection of $100\ \Omega$ and then finally into $70\ \Omega$

lines, as illustrated in Figure 6. This structure is designed using a full electromagnetic wave computer software CST, fabricated using LPKF D104 milling machine, and measured using Anritsu vector network analyzer (VNA). The fabricated prototype is shown in Figure 7.

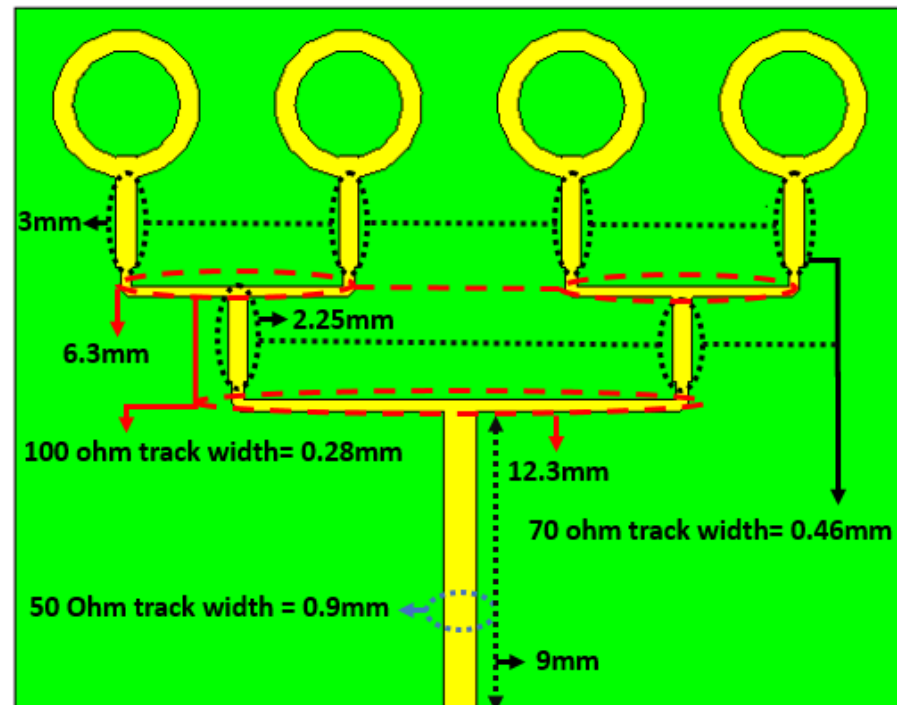


Figure 6. Linear array transformation of proposed antenna.

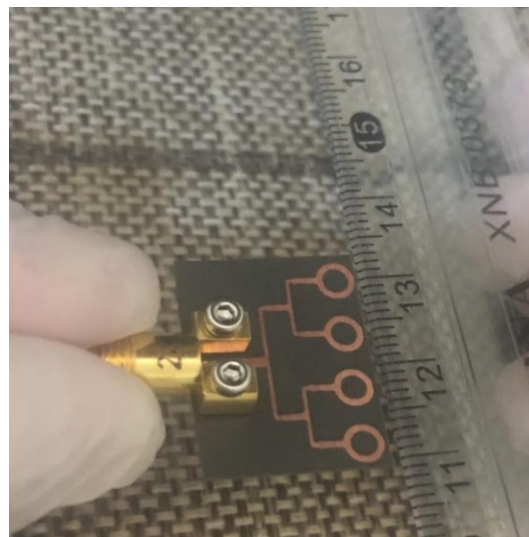


Figure 7. Fabricated Prototype.

3. Results and Discussion

A comparison between the experimental and the simulated reflection coefficient for the feed port is shown in Figure 8. The results obtained agree very well. However, minor inconsistencies occur, due to errors in fabrication, as well as the unavoidable use of coaxial cables during the measurement of the antenna. In Figure 9, the surface current distribution for the antenna system is illustrated at 28 GHz. Please note that the current is distributed symmetrically between the radiating elements.

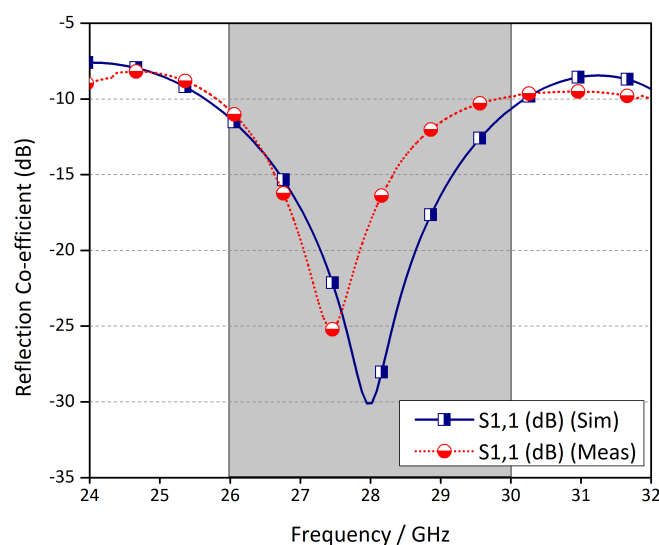


Figure 8. Simulated and Measured reflection co-efficient of proposed Antenna Array.

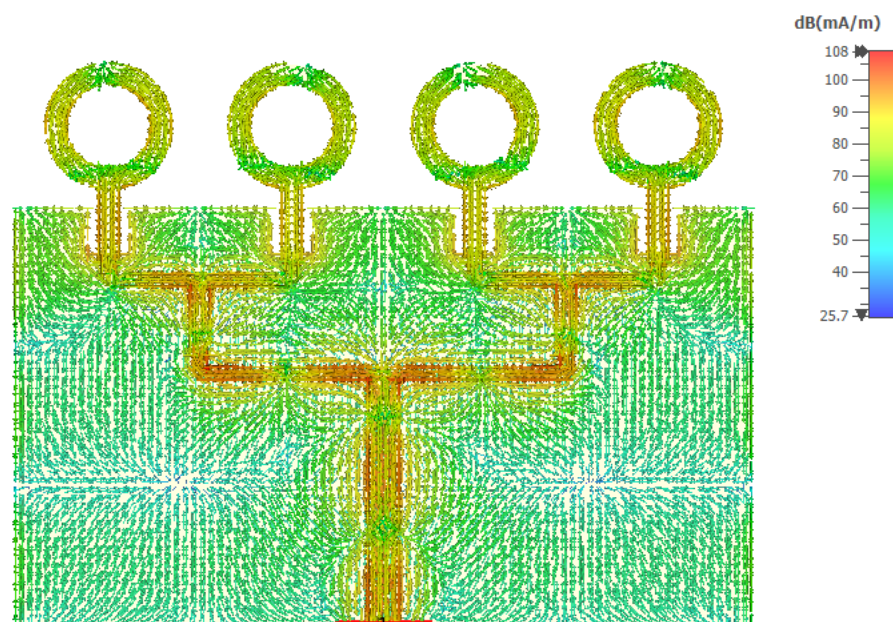


Figure 9. Surface Current Distribution of Array.

The simulated and the measured far-field fields of the system for $\phi = 0^\circ$ and $\phi = 90^\circ$ are shown in Figure 10. It is worth mentioning that the fields radiated by the antenna in yz -plane are almost round with a null at $\theta = \pm 90^\circ$. Similarly, the pattern in the xz -plane is directive with a beamwidth of 40° . This far-field profile suggests that it can be used for 5G technologies, where the front and the back with the same radiation characteristics are required. Moreover, a 3-D far-field pattern for the proposed system is depicted in Figure 11 for a perspective view. In addition, measured and simulated gain and efficiency are shown in Figure 12. The measured gain is more than 9 dBi within the desired frequency bandwidth, while the efficiency is around 90%. Therefore, one can conclude that with these attributes the proposed system will find its applications in the future 5G systems and subsystems.

To further demonstrate the advantages of this work with the available literature, a brief comparison in Table 2 has been drawn and it is observed that this work has advantage in terms of size and key performance parameters.

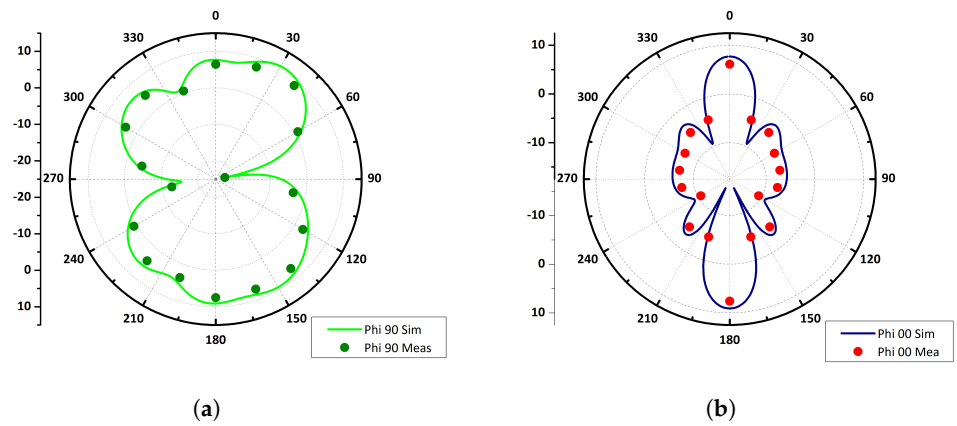


Figure 10. Radiation Patterns Simulated and Measured. (a) $\Phi = 90^\circ$. (b) $\Phi = 0^\circ$.

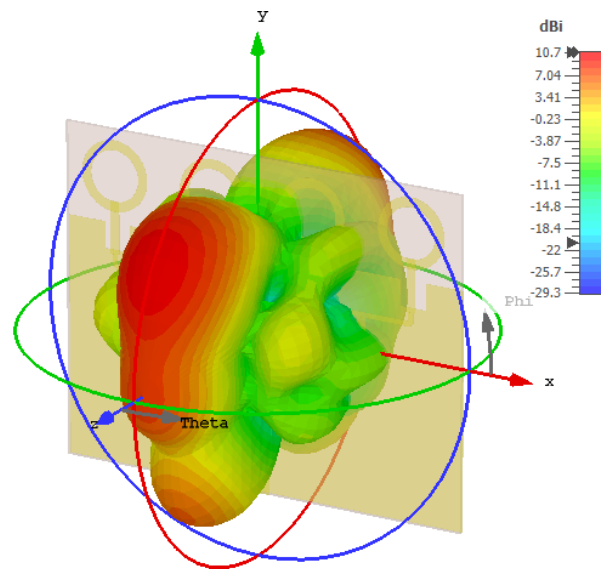


Figure 11. 3D Gain of proposed array at 28 GHz.

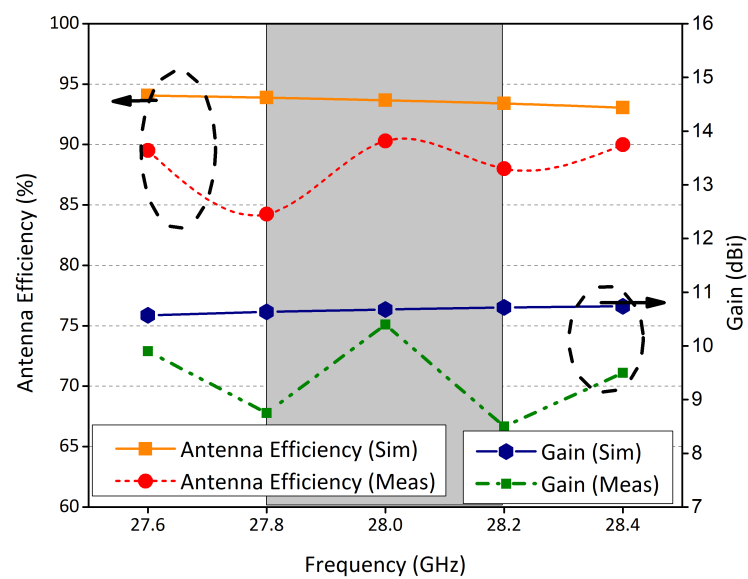


Figure 12. Performance parameter of proposed Array.

Table 2. Comparison Table.

Ref	Bandwidth (GHz)	Size L × W (mm)	Antenna Elements	Configuration	Gain (dBi)	Efficiency (%)
[5]	1.9	29.9 × 29.5	4	Planar	5.5	80
[6]	0.45	20 × 22	4	Planar	6.5	75
[7]	5	23 × 24	4	SIW	10	82
[9]	2.2	70 × 63.5	4	SIW	14	62
[16]	3	20 × 45	4	SIW	10.5	83
[17]	0.6	20 × 27	4	Planar	9	89
Proposed	4	20 × 22	4	Planar	10.2	90

4. Conclusions

In this work, a simple, efficient, low-cost, and easy to fabricate and integrate antenna system for 5G technologies is proposed. The proposed structure designed to resonate at central frequency of 28 GHz is composed of a ring shape patch with a square slot etched at the top mid-section of partial ground plane. The proposed structure is transformed into four element array with feed network for enhance performance characteristics. The proposed system has 90% efficiency, more than 10 dBi gain, and dual beam far-field profile. A prototype is fabricated to verify the simulated results and excellent agreement is observed between the measured and the computed results. Based on these attributes, it is believed that this system will find its applications in future 5G systems and subsystems.

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