

Article

A New Arrangement of Active Coils for Wireless Charging of UAV

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Abstract: This paper presents the design and optimization of a wireless power transfer (WPT) charging system based on magnetically coupled resonant technology, applied to an Unmanned Aerial Vehicle (UAV). In this paper, a charging system, including dual active transmitter coils and a single receiver coil, is proposed. The dual transmitting coils adopt a coaxial structure with different radii. This structure simplifies the calculation of the complex mutual inductance between the coils to a function of mutual inductance only related to the value of the radial misalignment. Aiming toward a constant charging power, the optimal transmission efficiency of electric energy is achieved by controlling the input voltages of the active coils, which are solved via a set of equations defined as Lagrange multipliers. The simulation results of the 570 V and 85,000 Hz system verified the validity of the proposed wireless UAV charging scheme.

Keywords: Lagrange multiplier; magnetically coupled resonant technology; power compensation; radial misalignment



Citation: He, Z.; Li, Z.; Wang, R.; Fan, Y.; Xu, M. A New Arrangement of Active Coils for Wireless Charging of UAV. *Energies* **2021**, *14*, 5754. <https://doi.org/10.3390/en14185754>

Academic Editor: José Gabriel Oliveira Pinto

Received: 21 July 2021
Accepted: 9 September 2021
Published: 13 September 2021

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1. Introduction

UAVs undoubtedly have high research potential and application value, but the flight time of UAVs restricts their development. This is because a high-energy density lithium battery, which is adopted to the UAV, only permits a flight time of about 20–40 min [1].

Thus, researchers began to research the application of magnetically coupled resonant technology to charge UAVs for solving this limitation [2–5]. This method creates a base station where the batteries can be automatically charged after landing [3]. The proposed magnetically coupled resonant system in [2] is mainly composed of two coils, namely, a single transmitting coil and a single receiving coil. The power transmission efficiency between the two coils is a function of the vertical distance between the two coils and the power supply. When the two coils are coaxial, the power transmission efficiency of the coil is the highest, and when the distance between the two coils is increased, the power transmission efficiency decreases. In the practical charging process of UAVs, the coaxial position of the receiving coil and the transmitting coil cannot be guaranteed, namely, radial misalignment, which will lead to a reduction in output power and charging time delay [6]. Therefore, it is necessary to research the power compensation problem of misaligned charging.

Scholars have adopted various methods to solve this problem [7–18]. In [7–11], a reconfigurable resonator and numerous transmission coil arrays with different structures have been used to improve transmission efficiency. A relatively complete mutual inductance model for analyzing magnetically coupled resonant technology was established. A planar transmitting resonator with an array of 2×2 using a single feeding loop was presented in [12]. This structure could maintain high transfer efficiency in a plane. However, the power transfer efficiency function determined by this structure is still two-dimensional

in essence. This leads to the need for complex control strategies to achieve the goal of constant output power, which is not worth it in practical application. In [13], two circular planar spiral coils were applied to both receive and transmit circuits of a WPT system. The main disadvantage of this configuration is that it reduces the load capacity of the UAV. In [14], an automatic landing procedure for UAV was proposed, which fundamentally reduces the possibility of radial misalignment. A WPT charging system with a movable transmitting coil was introduced in [15]. When the UAV lands, the mobile transmitting coil can automatically align with the receiving coil. Of course, it requires a high-precision positioning system. A lightweight wireless charging system was presented in [17]. This design can improve the load capacity of UAVs without affecting the operation of various equipment carried by UAVs. In [18], Noriaki Oodachi et al. have proposed a wireless power transmission system to solve this problem by using phase weights of transmission coil arrays. When the receiving coil is not coaxial with the transmitting coil, according to the direction between the transmission coil and the receiving coil, the transmission coil of the coil array is excited by the transmission circuit and the corresponding phase weight to achieve the purpose of power compensation. However, this method also requires a large number of coils, and the control algorithm and implementation are complex, which increase the costs.

The system proposed in this paper is simple and effective. In order to compensate the power reduction caused by radial misalignment, two coaxial active coils with different radii are used as the transmitting coils, and the compensation is realized by controlling the different input power of the two coils. When the output power of one transmitting coil decreases due to radial misalignment, the other transmitting coil with different radii that are still coaxial is applied to compensate the output power, so as to maximize the transmission efficiency of the system, which satisfies the Lagrange multiplier. In addition, by controlling the input voltage of the two active coils of the system, a power supply with arbitrary output power can be designed. This design simplifies the efficiency function from a function with two-dimensional variables to a function with only a radial variable, so there is no need for a complex control algorithm, which makes it more practical.

2. Proposed Wireless Charging of UAV System

This paper presents the design of a UAV charging system with magnetically coupled resonant technology, as shown in Figure 1. The transmitting coil uses double coaxial active coils, with the receiving coil and transmitting coils coupled by air. The vertical distance between the UAV and the charging pile is 10 cm. Both the double transmitting circuit and the single receiving circuit adopt the topology of inductance and capacitor in series. The whole system realizes power transmission through the full resonance of coil and capacitor.

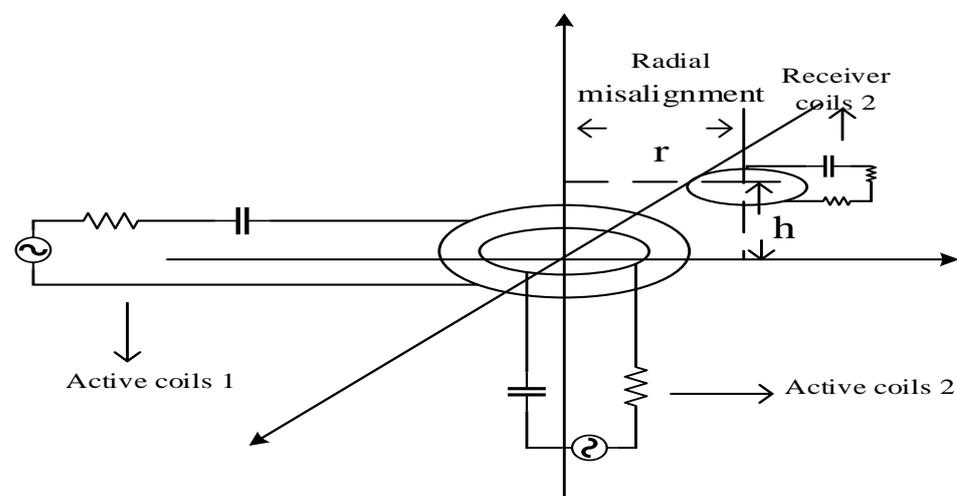


Figure 1. Proposed system for UAV charging.

In order to research the specific internal strategy of the new charging pile, two mathematical models, including the power model and mutual inductance model of the system, are established according to the actual charging situation of the UAV, and the theoretical analysis is carried out. The power model analyzes the relationship between the output power of the system and the input voltage of the two active coils; the transmission efficiency is found when the mutual inductance and output power are constant. The mutual inductance model analyzes the relationship between the mutual inductance of the system and the radial misalignment caused by the landing error when the UAV is charging, and provides a scheme to compensate for the system's output power reduction that is simple and effective. A 570 V, 85,000 Hz system is built in MATLAB/Simulink, and the output power of the system is obtained. The transmission efficiency of the system can reach 82%. At the same time, it is confirmed that the double transmitting coils have a good compensation effect on the power reduction caused by the radial misalignment of the system; the correctness of the control strategy is also verified.

2.1. Model of System Mutual Inductance and Radial Misalignment Distance

In most of the experimental studies, the transmitting coil and receiving coil are placed in parallel, and their central axis position is the same. However, when the UAV actually stops, it cannot be guaranteed that it will stop in the axial direction of the transmitting coil. When the position of the coil deviates, the mutual inductance M will change. When the system frequency is fixed and the coil resistance is constant, the change in mutual inductance M is the most important factor affecting system performance. Therefore, when the receiving coil has radial misalignment, the mutual inductance between the receiving coil and the transmitting coil changes, which leads to a change in system power and efficiency.

Only the mutual inductance mode of the single transmitting and single receiving coils is considered, as shown in Figure 2. The center of the transmitting coil is placed in the coordinate system $(0, 0, 0)$, and the center of the receiving coil is placed in the coordinate system $(0, t, h)$.

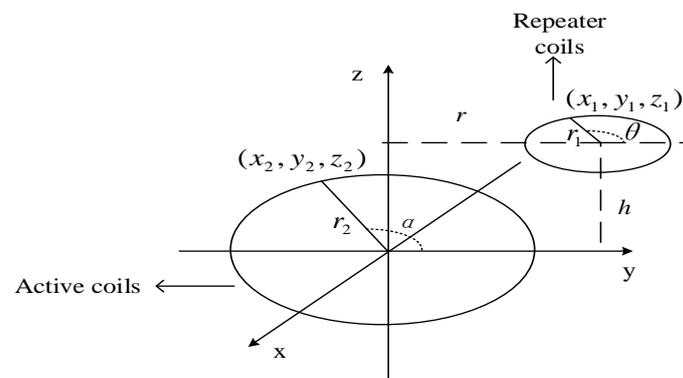


Figure 2. One active to receiver charging mode.

The parameter equations of the transmitting and receiving coils are listed as follows:

$$\begin{cases} x_1 = r_1 \times \cos \theta \\ y_1 = r_1 \times \sin \theta + r \\ z_1 = h \\ x_2 = r_2 \times \cos \alpha \\ y_2 = r_2 \times \sin \alpha \end{cases} \quad (1)$$

Therefore:

$$\begin{cases} dl_1 = (r_1 \times \cos \theta - r_1 \times \sin \theta) d\theta \\ dl_2 = (r_2 \times \cos \alpha - r_2 \times \sin \alpha) d\alpha \\ dl_1 \times dl_2 = r_1 \times r_2 \cos(\theta - \alpha) d\theta d\alpha \end{cases} \quad (2)$$

According to the Neumann formula, when the number of turns of the receiving coil is N_1 and the number of turns of the transmitting coil is N_2 , the mutual inductance formula between them is as follows:

$$M = \frac{N_1 \times N_2}{4\pi} \times \frac{dl_1 \times dl_2}{R} = \frac{N_1 \times N_2}{4\pi} \int_0^{2\pi} \int_0^{2\pi} \frac{r_1 \times r_2 \cos(\theta - \alpha) d\theta d\alpha}{D} \quad (3)$$

In the above formula:

$$D = \sqrt{(r_1 \times \cos \theta - r_2 \times \cos \alpha)^2 + (r_1 \times \sin \theta + r - r_2 \times \sin \alpha)^2 + h^2}$$

Analysis of Equation (3) shows that the mutual inductance of the system decreases with the increase in radial misalignment, and the output power decreases with the decrease in mutual inductance. Therefore, it is necessary to compensate for the mutual inductance reduction caused by the radial misalignment of the single transmitter and single receiver system.

Considering double transmitting and single receiving coils, the two transmitting coils are coaxial with different radii. With the increase in radial misalignment, the mutual inductance between the two transmitting coils and the receiving coil changes according to Equation (3). However, for the receiving coil, the total mutual inductance is the sum of mutual inductance between the two transmitting coils and the receiving coil, which compensates for the reduction in mutual inductance; Maxwell simulation verifies this conclusion.

2.2. Model of Output Power and Input Voltage

We define the mutual inductance between coil 1 and coil 2 as M_{12} , and define M_{13} and M_{23} in the same way. When the UAV lands on the charging pile, the relative position between the receiving coil and the two transmitting coils is fixed, so the mutual inductance between the two coils is fixed, i.e., M_{13} , M_{23} and M_{12} are constant. The output power of the system is only related to the input voltage of the active coil, as shown in Figure 3. The detailed model analysis is as follows:

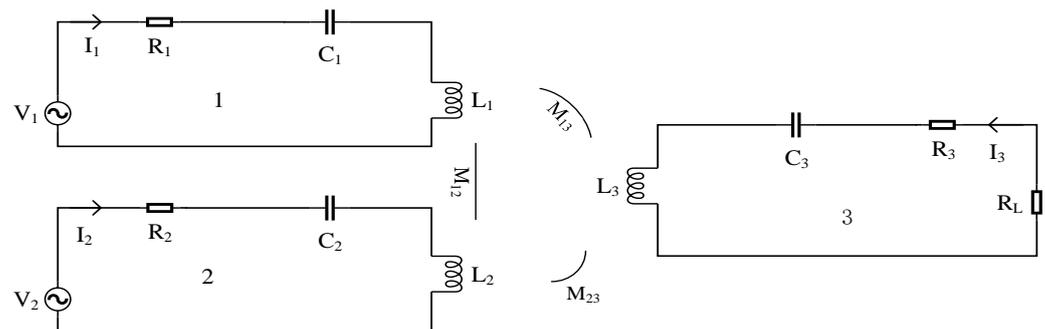


Figure 3. Two active to receiver charging modes.

In order to simplify the model, the parameters of the circuit model are symmetrical, namely, $R_1 = R_2 = R_3 = R$. The transmitter and receiver of the system will work at the same frequency, which is defined as:

$$\omega = 2\pi f = \frac{1}{\sqrt{C_1 L_1}} = \frac{1}{\sqrt{C_2 L_2}} = \frac{1}{\sqrt{C_3 L_3}} \quad (4)$$

where f is the fundamental frequency of the power supply.

By listing the voltage equation of each circuit, the following voltage and current (Equation (5)) can be obtained:

$$\begin{pmatrix} V_1 \\ V_2 \\ 0 \end{pmatrix} = \begin{pmatrix} R_1 + j\omega L_1 + \frac{1}{j\omega C_1} & -j\omega M_{12} & -j\omega M_{13} \\ -j\omega M_{12} & R_2 + j\omega L_2 + \frac{1}{j\omega C_2} & -j\omega M_{23} \\ -j\omega M_{13} & -j\omega M_{23} & R_3 + R_1 + j\omega L_3 + \frac{1}{j\omega C_3} \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \\ I_3 \end{pmatrix} \quad (5)$$

The input impedance of the system is solved. According to the defined system operating frequency, order $V_2 = 0$, the following equation (Equation (6)) can be obtained:

$$Z_{\text{sent1}} = R_1 + \frac{(\omega M_{12})^2(R_3 + R_1) - j\omega^3 M_{12} M_{13} M_{23}}{R_2(R_3 + R_1) + (\omega M_{23})^2} + \frac{(\omega M_{13})^2 R_2 + j\omega^3 M_{12} M_{13} M_{23}}{R_2(R_3 + R_1) + (\omega M_{23})^2} = R_1 + Y_2 + Y_3 \quad (6)$$

Y_2 and Y_3 in Equation (6), respectively, indicate the mutual inductance influence of loop 2 and loop 3 on loop 1:

$$\begin{cases} Y_2 = \frac{(\omega M_{12})^2(R_3 + R_1) - j\omega^3 M_{12} M_{13} M_{23}}{R_2(R_3 + R_1) + (\omega M_{23})^2} \\ Y_3 = \frac{(\omega M_{13})^2 R_2 + j\omega^3 M_{12} M_{13} M_{23}}{R_2(R_3 + R_1) + (\omega M_{23})^2} \end{cases} \quad (7)$$

Therefore, the input current I_1 of active coil 1 can be expressed as:

$$I_1 = \frac{V_1}{Z_{\text{sent1}}} = \frac{V_1}{R_1 + Y_2 + Y_3} \quad (8)$$

Aligned, the input current I_2 of active coil 2 can be expressed as:

$$I_2 = \frac{V_2}{Z_{\text{sent2}}} = \frac{V_2}{R_2 + Y_2' + Y_3'} \quad (9)$$

Among:

$$\begin{cases} Y_2' = \frac{(\omega M_{12})^2(R_3 + R_1) - j\omega^3 M_{12} M_{13} M_{23}}{R_1(R_3 + R_1) + (\omega M_{13})^2} \\ Y_3' = \frac{(\omega M_{23})^2 R_1 + j\omega^3 M_{12} M_{13} M_{23}}{R_1(R_3 + R_1) + (\omega M_{13})^2} \end{cases} \quad (10)$$

Considering two active coils, the output current I_3 of the receiver can be expressed as:

$$I_3 = \frac{j\omega M_{13} \times I_1 + j\omega M_{23} \times I_2}{R_3 + R_1} = \frac{j\omega M_{13}}{R_3 + R_1} \times \frac{V_1}{R_1 + Y_2 + Y_3} + \frac{j\omega M_{23}}{R_3 + R_1} \times \frac{V_2}{R_2 + Y_2' + Y_3'} \quad (11)$$

Among $Z_{\text{sent2}} = R_1 + Y_2' + Y_3'$. Therefore, the input power P_{in} and output power P_{out} of the system can be expressed as:

$$P_{\text{in}} = \text{Re}(V_1 \times I_1^*) + \text{Re}(V_2 \times I_2^*) = \frac{(V_1)^2}{R_1 + Y_2 + Y_3} + \frac{(V_2)^2}{R_2 + Y_2' + Y_3'} \quad (12)$$

$$P_{\text{out}} = I_3^2 \times R_1 = \left(\frac{j\omega}{(R_3 + R_1)} \right)^2 \times \left(\frac{M_{13} V_1}{R_1 + Y_2 + Y_3} + \frac{M_{23} V_2}{R_2 + Y_2' + Y_3'} \right)^2 \times R_1 \quad (13)$$

Therefore, the efficiency of the system η can be calculated as follows:

$$\eta = \frac{P_{\text{out}}}{P_{\text{in}}} = \left(\frac{j\omega}{R_3 + R_1} \right)^2 \times \frac{Z_{\text{sent1}} \times Z_{\text{sent2}} \times \left(\frac{M_{13} V_1}{Z_{\text{sent1}}} + \frac{M_{23} V_2}{Z_{\text{sent2}}} \right)^2}{(V_1)^2 \times Z_{\text{sent2}} + (V_2)^2 \times Z_{\text{sent1}}} \times R_1 \quad (14)$$

Thus, the value of η is related to the two input voltages of the system and the mutual inductance between the two coils. Additionally, the observation Equations (13) and (14) show that when the output power and radial misalignment are fixed, there must be the maximum transmission efficiency.

The output power P_{out} is set as constant at 640 W. The goal is to maximize transmission efficiency. The optimal problem formulation is Equation (15):

$$\left\{ \begin{array}{l} \text{Max} : \left(\eta = \frac{P_{out}}{P_{in}} = \left(\frac{j\omega}{R_3 + R_1} \right)^2 \times \frac{Z_{sent1} \times Z_{sent2} \times \left(\frac{M_{13}V_1}{Z_{sent1}} + \frac{M_{23}V_2}{Z_{sent2}} \right)^2}{(V_1)^2 \times Z_{sent2} + (V_2)^2 \times Z_{sent1}} \times R_1 \right) \\ P_{out} - 640 = 0 \end{array} \right\} \quad (15)$$

Lagrange multiplier λ is introduced and a Lagrangian function is constructed:

$$L = \eta - \lambda \times (P_{out} - 640) \quad (16)$$

When the partial derivative of Equation (16) is found, they can be set to zero:

$$\left\{ \begin{array}{l} \frac{\partial L}{\partial V_1} = \frac{\partial \eta}{\partial V_1} - \lambda \times \frac{\partial P_{out}}{\partial V_1} = 0 \\ \frac{\partial L}{\partial V_2} = \frac{\partial \eta}{\partial V_2} - \lambda \times \frac{\partial P_{out}}{\partial V_2} = 0 \\ P_{out} = 640 \end{array} \right. \quad (17)$$

The corresponding V_1 and V_2 are obtained when the efficiency is maximum, as long as V_1 and V_2 are satisfied (Equation (17)).

The flowchart of the methodology mentioned is introduced in Figure 4.

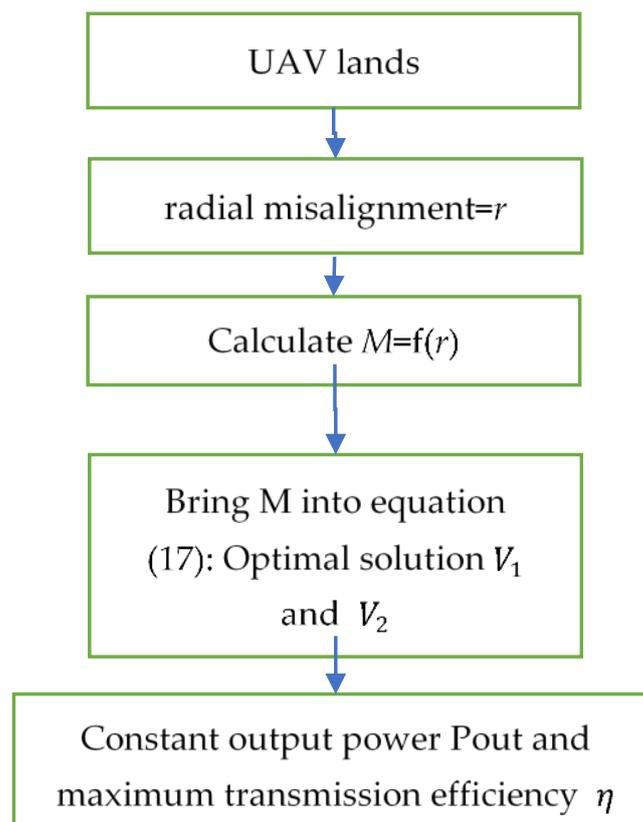


Figure 4. The flowchart of the methodology.

3. Simulation

The proposed 570 V and 85,000 Hz power model is simulated in MATLAB/Simulink, and the rationality of the model is verified. The parameters of the system are listed in Table 1. The mutual inductance parameters of the coil are simulated by the finite element method in Maxwell.

Table 1. Parameters of the system.

Parameter	Value	Parameter	Value
L_1 & L_3	500.0 μH	Peak amplitude of V_1 & V_2	570 V
C_1 & C_3	0.0072 μF	f	85,000 Hz
R_1 & R_2 & R_3	0.2 Ω	M_{13}	175 μH
R_L	20 Ω	M_{23}	300 μH
M_{12}	400 μH	C_2	0.0036 μF
L_2	1000 μH		

3.1. Simulation of Mutual Inductance Parameters

When the coil has radial misalignment, double transmitting and single receiving coils are considered. The radii of the two transmitting coils are 26 cm and 36 cm, respectively, and the radius of the receiving coil is 26 cm. The vertical distance between the transmitting coil and the receiving coil is kept at 10 cm, and the medium is air. The finite element simulation, as shown in Figure 5a, is carried out in Maxwell. The relationship between mutual inductance and radial distance in the two transmitting coils and receiving coils can be obtained, as shown in Figure 5b,c. It is found that the mutual inductance between the two transmitting coils and the radial distance decreases with the increase in radial misalignment. Combined with the power model, the power of the system also decreased. For the receiving coil, mutual inductance is the sum of the two, which compensates for the reduction in mutual inductance and power of the single transmitting coil. At the same time, the self-inductance of the 36 cm coil is twice that of the other two 26 cm coils, and the self-inductance is only related to the material and radius of the coil.

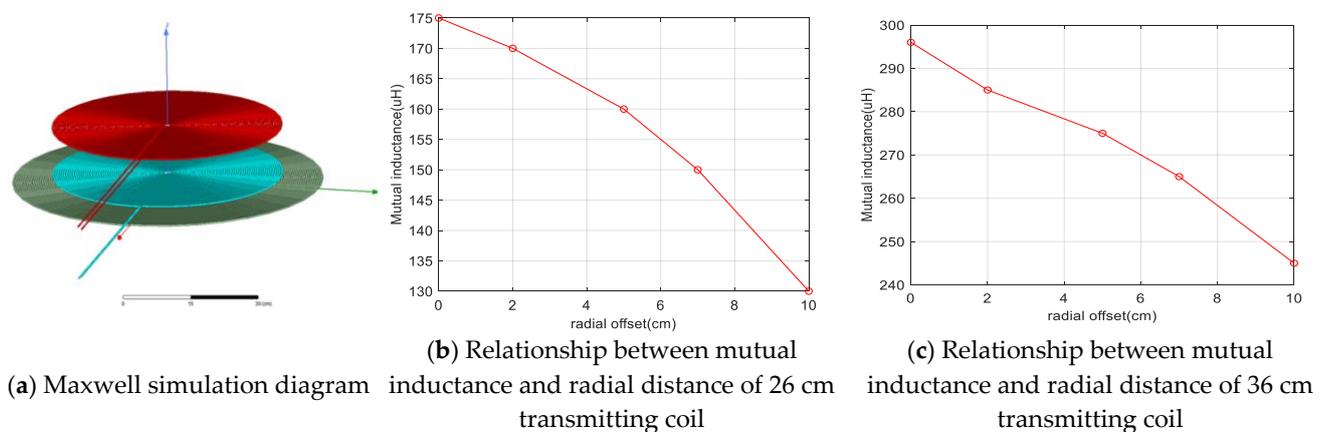
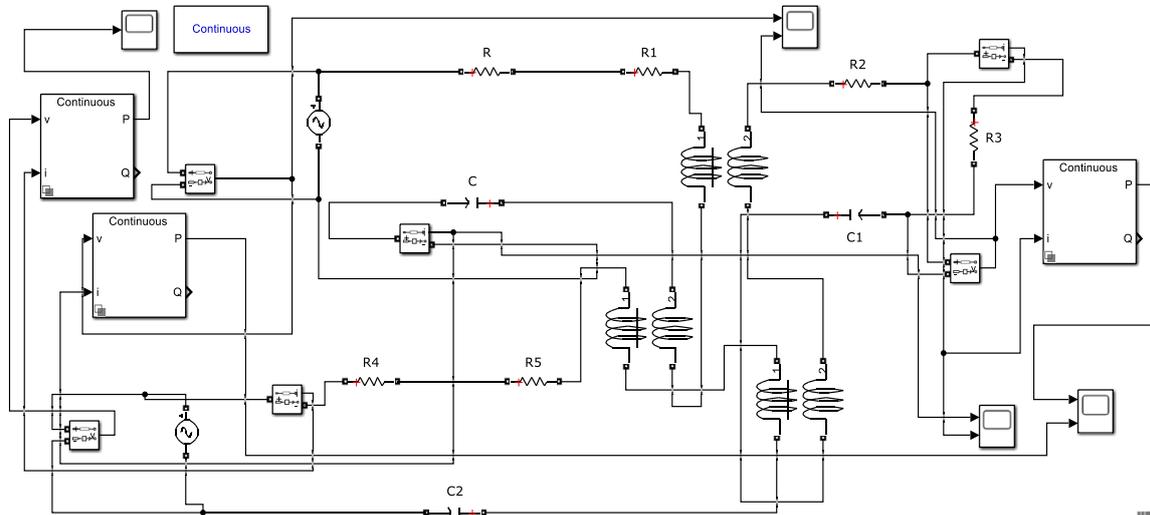


Figure 5. Maxwell simulation. (a) Snapshot of Maxwell simulation diagram (b) Simulation result of 26 cm transmitting coil. (c) Simulation result of 36 cm transmitting coil.

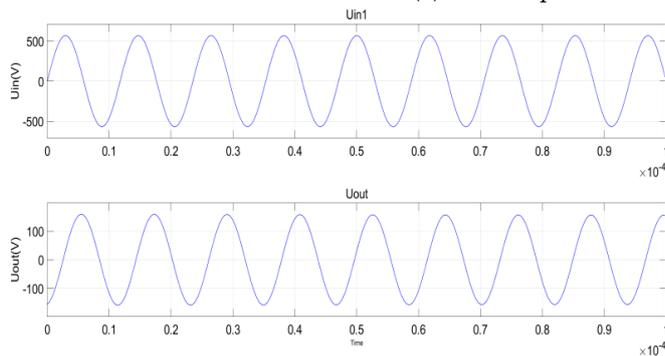
3.2. Power

According to the coil parameters calculated in model A, the simulation output waveform of the system simulated by Simulink is shown in Figure 6. A snapshot of the simulation file is shown in Figure 6a. We use a transformer to replace the resonant coils and their mutual inductance. As for the high-frequency AC power required by the system, we directly use AC power to generate it. Both the double transmitting circuit and single receiving circuit adopt the topology of inductance and capacitor in series. In other figures, U_{out} is the output voltage waveform. U_{in} is the input voltage of the active coil. I_{in} is the input current waveform of the active coil. I_{out} is the output current waveform. Output power is the output active power of the system and input power1 and input power2 are the input active power of the system. According to the simulated voltage and current, the transmission efficiency of the system can be calculated, which reaches 94%. Comparing the simulation results with the theoretical results, the output power error is less than 5%.

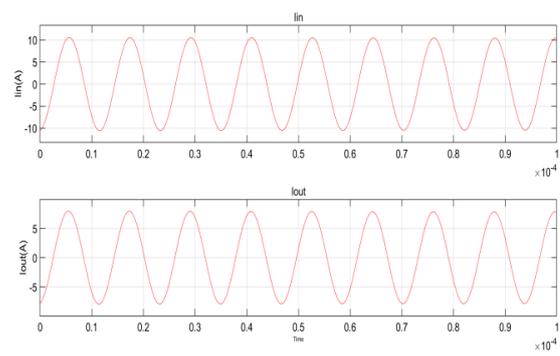
At the same time, the simulation results show that the output current and voltage waveform is sine wave, but there are phase differences, which is caused by mutual inductance resonance not being complete, and therefore, there is phase lag.



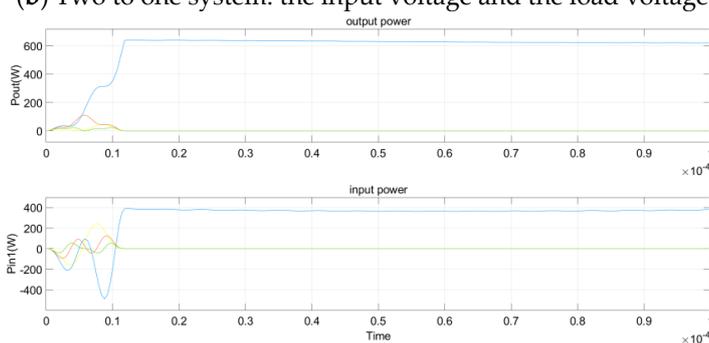
(a) One snapshot of the simulation file



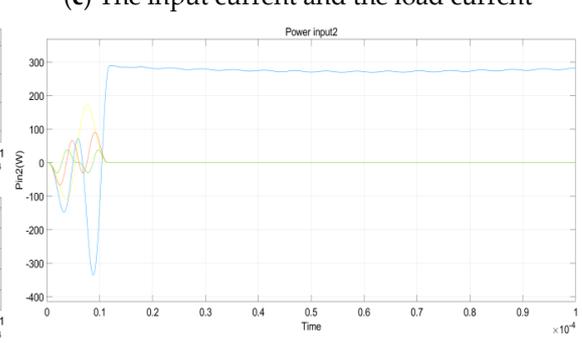
(b) Two to one system: the input voltage and the load voltage



(c) The input current and the load current



(d) The output power and the input power1



(e) The input power2

Figure 6. Waveforms of the transmitter currents and the load voltage. (a) Snapshot of Matlab simulation (b) Voltage (c) Current (d) Output power and input power 1 (e) input power 2.

By changing the value of the input voltage, an arbitrary power output can be achieved. Table 2 shows the simulated output power under different voltage peaks, which is close to the theoretically derived value.

In addition, different radial misalignment will cause a decrease in the mutual inductance between the receiving coil and the transmitting coil, resulting in a decrease in power. As shown in Figure 7, when the peak input voltages of the two active coils are both 570 V, the output power is negatively correlated with the radial misalignment.

Finally, when the radial misalignment is constant with the input voltage of one active coil, the simulation between the output power and the input voltage of the other active coil verifies the correctness of the control strategy.

The verification result is shown in Figure 8. At this time, the radial misalignment of the system is 5 cm, and the input voltage of an active coil is 570 V. The output power of the system increases as the input voltage of the active coil increases.

Table 2. Comparison of simulation and theoretical values of output power.

	Value	Value	Value	Value	Value
Peak Amplitude of Input Voltage ($V_1 = V_2$) (V)	300	400	500	570	600
Simulation (W)	180	310	500	640	710
Theoretical (W)	186	320	525	662	739

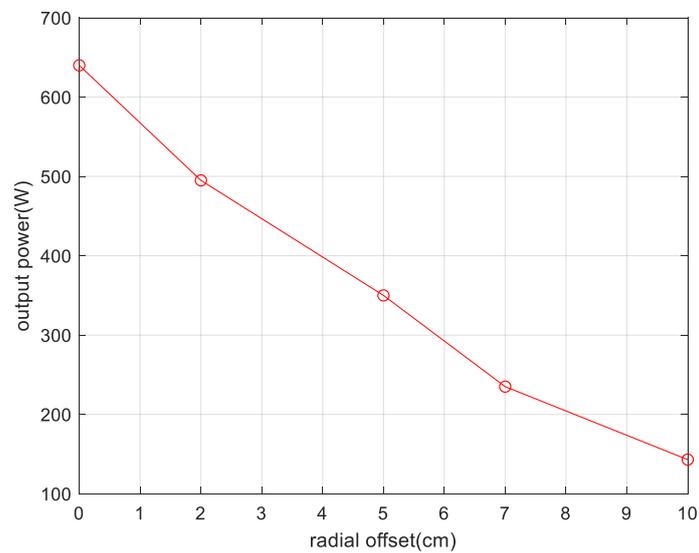


Figure 7. The output power and the radial misalignment.

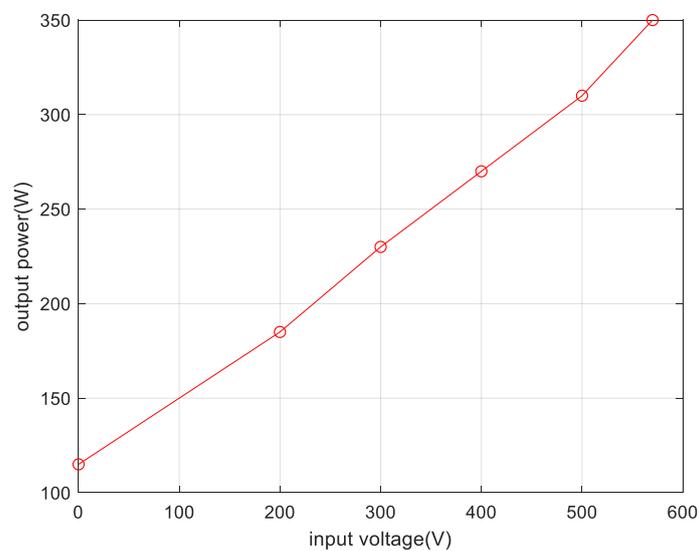


Figure 8. The output power and the input voltage.

4. Discussion

This paper presents a new wireless power transmission charging system for UAVs, and establishes two mathematical models to perfect the theoretical basis of the system. We also determine the control strategy between output power and input voltage when there is radial misalignment. The structure of a double transmitting coil and single receiving coil can better compensate for the mutual inductance and output power drop caused by radial misalignment. At the same time, the output power of the system can be changed by controlling the input voltage of the system to realize a power supply of any output power, which is simple and easy to operate. When the output power is constant, maximum power transfer efficiency is obtained by controlling the input voltage of the two coils. Numerical and simulation results have been presented in terms of electrical performance to demonstrate the validity of the proposed structure.

Author Contributions: Conceptualization, Z.H. and Z.L.; methodology, Z.H.; software, Z.H.; validation, M.X., Y.F. and R.W.; formal analysis, Z.H.; investigation, M.X.; resources, Z.L.; data curation, R.W.; writing—original draft preparation, Z.H.; writing—review and editing, Z.H.; visualization, R.W.; supervision, Z.L.; project administration, Z.H.; funding acquisition, Z.L. All authors have read and agreed to the published version of the manuscript.

Funding: This research received no external funding.

Data Availability Statement: The data presented in this study are available on request from the corresponding author.

Acknowledgments: Thanks are due to Wenjing Xiong and Zixi Liu, of Central South University, for their help in the testing.

Conflicts of Interest: The authors declare no conflict of interest.

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