



A Hollow-Core Photonic-Crystal Fiber-Optic Gyroscope Based on a Parallel Double-Ring Resonator

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Abstract: A novel system structure of resonant fiber optical gyroscope using a parallel double hollowcore photonic crystal fiber ring resonator is proposed, which employs the double closed loop and reciprocal modulation–demodulation technique to solve the problem of the length mismatch between rings. This structure can suppress the residual amplitude modulation noise and laser frequency noise, essentially eliminating the influence of the Rayleigh backscattering noise and dramatically reduce the Kerr-effect-induced drift by three orders of magnitude. Thanks to its excellent noise suppression effect, the sensitivity of this novel system can approach the shot-noise-limited theoretical value of 8.94×10^{-7} rad/s assuming the length of the fiber ring resonator is 10 m.

Keywords: fiber optic sensor; resonant fiber optic gyroscope; parallel double-ring resonator; hollow-core photonic-crystal fiber; Rayleigh backscattering noise; Kerr-effect-induced drift

1. Introduction

The resonant fiber optic gyroscope (RFOG) using a high-coherence light source can achieve high precision angular velocity measurement by detecting the resonance frequency difference caused by the Sagnac effect [1]. The RFOG has the potential to offer the same performance rotation sensing with a significantly shortened length compared with the commercial interferometric fiber-optic gyroscope (IFOG) [2]. This size advantage makes the RFOG very attractive for inertial navigation application. However, there are many parasitic effects, including Rayleigh backscattering noise [3], Kerr-effect-induced drift [4], temperature-driven polarization instability [5], residual amplitude modulation (RAM) noise [6], and laser frequency noise [7], deteriorating the performance of RFOG and making it far beyond the shot-noise-limited theoretical sensitivity.

Various methods have been reported for eliminating these parasitic effects. The Rayleigh backscattering noise can be reduced by phase modulators in both launching paths driven by different modulation frequencies, and using the phase modulators to suppress the carrier [3,8,9]. However, the use of separate modulation frequencies for clockwise (CW) and counterclockwise (CCW) waves, means that second harmonic distortion and RAM [6,9,10] will not be reciprocal to CW and CCW waves. Since these noises cause errors in the determination of the resonance center, they consequently translate to bias errors. The Kerr-effect-induced drift results from the light intensity difference between the CW and CCW waves circulating in the fiber ring resonator (FRR), which can be diminished by implementation of the light intensity feedback control technique [11–13] or adoption of spun fiber [14]. But research on these schemes have been restricted to complex structures that are not suitable for miniaturization. As for the drift caused by temperature-driven polarization instability, other groups have successfully reduced much of this drift by adding dual 90° polarization axis rotated splices [15] or adoption of a single-polarization resonator based on a micro-optical polarizing coupler [16]. These techniques require



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Copyright: © 2021 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https:// creativecommons.org/licenses/by/ 4.0/). extremely precise assembly or difficult splices. Besides, splices or other polarizationdependent devices add undesirable loss and backscatter to the FRR. The emergence of hollow-core photonic-crystal fiber (HCPCF) offers a novel approach to reduce the Kerreffect-induced drift [17,18] and polarization errors [19,20]. However, other experimental results indicated that the backscattering coefficient of some commercial HCPCFs are two orders of magnitude lager than the typical backscattering coefficient observed for SMF-28 fiber [21,22], which means the RFOG with a single HCPCF ring resonator may increase the Rayleigh backscattering noise.

From above, we find that these parasitic effects have been addressed by competing, or mutually exclusive countermeasures, and there is no countermeasure that can solve all these problems simultaneously. In this paper, we propose a novel system structure of RFOG using a parallel-double HCPCF ring resonator, collectively referred to as PDHC-RFOG. First, the Rayleigh backscattering noise in the PDHC-RFOG is analyzed in-depth. In addition, we establish a theoretical model to analyze the angular velocity drift induced by the Kerr effect in the PDHC-RFOG. Numerical simulations show that the Rayleigh backscatters in the PDHC-RFOG will not yield the demodulation error at the gyro output, and the PDHC-RFOG can reduce the Kerr-effect-induced drift by three orders of magnitude compared to the conventional RFOG using a single ring resonator. Moreover, the PDHC-RFOG is expected to suppress the temperature-driven polarization instability by a factor of 10 [20]. Applying the double closed loop [23] and reciprocal modulation–demodulation technique [10] to the PDHC-RFOG, it is encouraging that other parasitic performance barriers in the conventional RFOG, such RAM noise and laser frequency noise, can also be dramatically reduced.

2. The PDHC-RFOG System

A schematic diagram of the PDHC-RFOG employing the double closed loop [23] and reciprocal modulation-demodulation technique is depicted in Figure 1. The light emitted by the high-coherence laser is split into two equal-power beams (CW waves, CCW waves) by the LiNbO3 Y-branch phase modulator (Y-PM), and modulated with the same modulation frequency f_m . The phase modulator 1 (PM1) is driven by a seriodyne wave for frequency shifting and PM2 is used for the same purpose [24]. Then the CW and CCW waves enter two HCPCF ring resonators through a 2 \times 2 coupler, where fiber ring resonator 1 (FRR1) transmits CW waves, while ring resonator 2 (FRR2) transmits the CCW waves. A series of micro-optical lenses can be employed in the PDHC-RFOG as the 2×2 coupler, thus avoiding the polarization crosstalk in the resonator input coupler for conventional polarization-maintaining fiber [16,19]. Therefore, the light waves that originally propagated in the same FRR are changed to propagate separately in two independent HCPCF ring resonators without affecting each other. The CW and CCW waves from two HCPCF resonators are detected by photodetectors, PD1 and PD2, respectively. The output of PD1 is fed back through lock-in amplifier 1 (LIA1) and Frequency Feedback 1 to lock the laser frequency on the CW resonant frequency, which constitutes the first closed loop of the PDHC-RFOG. The output of PD2 is fed back through lock-in amplifier 2 (LIA2) and Frequency Feedback 2 to lock the laser frequency on the CCW resonant frequency, which constitutes the second closed loop of the PDHC-RFOG. Frequency Feedback 1 generates a serrodyne wave with an equivalent frequency shift of Δf_{cw} . Similarly, Frequency Feedback 2 generates a serrodyne wave with an equivalent frequency shift of Δf_{ccw} . Finally, the difference between two demodulation outputs gives the output of the rotation rate.



Figure 1. Schematic diagram of the PDHC-RFOG system.

Taking the CCW waves as an example, the output intensity of the FRR2 normalized by the input intensity can be derived as:

$$H_{ccw} = e^{-\alpha_c} \left[1 - \rho \frac{(1-Q)^2}{(1-Q)^2 + 4Qsin^2(\omega\tau/2)} \right],$$
(1)

where

$$\begin{cases} \omega = 2\pi f_0 \\ \rho = 1 - \frac{\left[T - (TQ + R)\right]^2}{e^{-\alpha_c}(1 - Q)^2} \\ T = e^{-\alpha_c/2}C_{bar} ; \\ R = C_{cross}^2 e^{-\alpha_c} e^{-\alpha_f L/2} \\ Q = C_{bar} e^{-\alpha_c/2} e^{-\alpha_f L/2} \end{cases}$$

where f_0 is the laser frequency; C_{bar} and C_{cross} are the coupling coefficients of the throughout arm and cross arm of the coupler, respectively; α_f and α_c are the additional loss of the fiber ring and coupler, respectively; L is the length of the FRR; $\tau = n_0 L/c$ is the transit time of the FRR, where n_0 is the refractive index of the fiber, and c is the speed of light in a vacuum.

The shot noise limited sensitivity is an import parameter of the RFOG, which can be expressed as [2,25]:

$$\Delta \Omega \approx \frac{\sqrt{2}\lambda_0 c}{LDF} \sqrt{\frac{hf_0}{\eta t_i I_0}} \frac{\sqrt{H_{max}}}{H_{max} - H_{min}},$$
(2)

where λ_0 is the wavelength of the laser source, *D* is the diameter of the ring resonator; *F* is the finesse of the ring resonator; η is the photodetector quantum efficiency; t_i is the integration time; I_0 is the input optical power corresponding to the input electric field E_0 ; *h* is Planck's constant; and H_{max} and H_{min} represent the maximum and minimum of the resonance curve, respectively. Therefore, based on the Equations (1) and (2), and the parameters in Table 1, the shot-noise-limited theoretical sensitivity of the PDHC-RFOG is calculated to be 8.94×10^{-7} rad/s.

Parameter	PDHC-RFOG
λ	1550 nm
L	10 m
D	0.1 m
n_0	1
$\alpha_{ m BS}$	$2.5 imes 10^{-5}$
$C_{\rm bar}$	$\sqrt{0.95}$
C _{cross}	$\sqrt{0.05}$
$lpha_c$	1 dB
α_f	20 dB/km *
I_0	1 mW
η	0.85
t_i	1 s

Table 1. Parameters for the PDHC-RFOG.

* Value provided by NKT Photonics.

It is important to note that the FRR1 and the FRR2 should be the same type of HCPCF, and the placement of the ring resonator, ring diameter and initial length must be consistent as far as possible. However, it is difficult to have exactly the same length for both rings in practical applications. Therefore, we adopt the double closed-loop scheme to solve the inherent nonreciprocity problem of the PDHC-RFOG. Assuming that the length of the two rings is 10 m and the length mismatch between rings is controlled to within the order of one centimeter, thus the difference in the free spectral range between rings is calculated to be about 29.97 kHz. For the serrodyne modulation technique [24], only a serrodyne wave with a period of 33.367 μ s is needed to enable the resonance of the light waves in the two rings, which can be realized theoretically.

Furthermore, the PDHC-RFOG using a double ring resonator structure allows both CW and CCW waves modulate by Y-PM with the same frequency, since the Rayleigh backscattering noise is naturally eliminated. In addition, this reciprocal modulation–demodulation technique makes the two counter-propagating light waves have the same RAM noise and laser frequency noise. Hence, the influence of these noises is eventually eliminated by differential output.

3. Suppression of the Rayleigh Backscattering Noise by the PDHC-RFOG

The Rayleigh backscattering noise limits the accuracy of the conventional RFOG. Although it has been proposed that the Rayleigh backscattering noise may be reduced by using HCPCFs since most of the mode travels through air instead of silica, but there were some measurements confirming that the backscattering coefficient for some commercial HCPCFs are still too high [21,22] due to dimensional fluctuations in the manufacturing process. Hence, the Rayleigh backscattering noise increases for conventional RFOGs with a single HCPCF ring resonator.

The PDHC-RFOG we proposed can essentially eliminate the influence of the Rayleigh backscatters. We analyzed the Rayleigh backscatters in the CCW direction of the conventional RFOG with a single-ring resonator and the PDHC-RFOG, respectively. Figure 2a shows the Rayleigh backscattering noise in the conventional RFOG, as the backscattered waves from CW waves encounter and interfere with CCW waves, the bias errors result. In addition, CW backscattered waves also cause errors from their own light intensity falling on the wrong photodetector.



Figure 2. A simplified model for analyzing the Rayleigh backscattering noise in the CCW direction of the RFOG: with (**a**) a single HCPCF ring resonator; with (**b**) the parallel-double HCPCF ring resonator.

The Equation (1) is applicable to both the conventional RFOG and the PDHC-RFOG. Moreover, as shown in the Figure 2a, the backscattered electric field of CW waves is given by

$$E_{b-cw} = \frac{E_0 C_{cross}^2 e^{-\alpha_c} e^{-\alpha_f L} \sqrt{\alpha_{BS}} e^{i(\omega t - 2\beta L - \theta)}}{\left(1 - Q e^{-i\omega\tau}\right)^2},$$
(3)

where α_{BS} is the Raleigh backscattering coefficient of fiber ring and θ is the phase difference between incident light waves and backscattered light waves.

Thus, in the conventional RFOG, the total electric field output is obtained as the sum of the Rayleigh backscattering, E_{b-cw} , and the CCW waves, E_{ccw} , which travel in the same direction as the backscatter, E_{b-cw} . Then, the total intensity output, $I_{out-single}$, is derived as

$$I_{out-single} = \left\langle \left(E_{ccw} + E_{b-cw}\right) \cdot \left(E_{ccw} + E_{b-cw}\right)^* \right\rangle = I_{ccw} + I_{b-cw} + I_i, \quad (4)$$

where the first term in Equation (4) is the CCW waves intensity given by Equation (1), the second term is the interference intensity between CCW waves and backscattered waves, and the last term is the backscattered waves intensity.

In contrast, in the PDHC-RFOG, the total electric field output only contains the CCW waves, E_{ccw} , because the CW waves propagate in another HCPCF ring resonator, and there are no backscatters traveling in the same direction as CCW waves. Then, the total intensity output, $I_{out-double}$, is given by

$$I_{out-double} = \langle E_{ccw} \cdot E_{ccw}^* \rangle = I_{ccw}.$$
⁽⁵⁾

The simulation parameters are listed in Table 1. From the static simulation result in Figure 3a, it is clear that I_{b-cw} (red dash line) and I_i (red dash-dotted line) have similar resonance characteristics to the CCW output, $I_{out-single}$ (green line). Therefore, the CCW output, $I_{out-single}$, under the influence of the intensity term, I_{b-cw} , and the interference term, I_i , result in (1) reduction of the signal-to-noise ratio and (2) change of the resonance depth, ρ_1 . Even worse, the gyroscope output will fluctuate randomly due to the phase term, θ , not being constant in practical applications [26]. In contrast, Figure 3b shows the CCW output in the PDHC-RFOG, the two counter-propagating waves, $I_{out-double}$ (green line) and I_{b-ccw} (green dash line), will not affect each other. Thus, only the CCW waves fall on the photodetector. Comparing Figure 3a and b, the resonance depth of the PDHC-RFOG ρ_2 is 1.14% above ρ_1 .



Figure 3. Simulated results of the Rayleigh backscattering noise in the CCW direction of the RFOG: with (**a**) a single HCPCF ring resonator; with (**b**) the parallel-double HCPCF ring resonator.

4. Suppression of the Kerr-Effect-Induced Drift by the PDHC-RFOG

The Kerr effect is one of the non-reciprocal optical effects that seriously deteriorates the performance of the RFOG. In the conventional RFOG, coherent optical propagation causes a large optical power density in the resonator owing to the small fiber diameter, as well as a significant change in the refractive index due to self-phase modulation (SPM) and cross-phase modulation (XPM) [4,27]. As shown in Figure 4, when the power of the CW waves and CCW waves in the FRR is unbalanced, a small non-reciprocal phase difference appears, which is superimposed on the non-reciprocal Sagnac phase difference causing a pseudo rotation drift signal, and eventually deteriorating the detection accuracy of the RFOG system.



Figure 4. CW and CCW waves in the parallel-double HCPCF ring resonator.

The PDHC-RFOG we proposed is not affected by the XPM effect, but only by the SPM effect because the counter-propagating light waves are propagating in two separate HCPCF rings. Therefore, the perturbation of propagation constants β_{kcw} and β_{kccw} caused by the Kerr effect into two counter-propagating light waves is only related to its own light intensities I_{cw} and I_{ccw} :

$$\beta_{kcw} = \frac{2\omega Z n_2}{c A_{eff}} (I_{cw})$$

$$\beta_{kccw} = \frac{2\omega Z n_2}{c A_{eff}} (I_{ccw})$$
(6)

where n_2 is the nonlinear refractive index coefficient caused by the Kerr effect in the fiber; A_{eff} is effective area of the propagation mode in the fiber; Z is the impedance of the fiber; u_{cw} and u_{ccw} are the power-splitting coefficients of two counter-propagating waves divided by the Y-PM. The input electric field E_{in_cw} and E_{in_ccw} with finite temporal coherence and phase modulation [4] are expressed by Equation (7)

$$E_{in_cw} = \sqrt{u_{cw}} \cdot E_0 e^{j\omega t} e^{j\phi(t)} e^{jM_{cw} sin(\omega_{cw} t)}$$

$$E_{in_ccw} = \sqrt{u_{ccw}} \cdot E_0 e^{j\omega t} e^{j\phi(t)} e^{jM_{ccw} sin(\omega_{ccw} t)} ,$$
(7)

where the phase fluctuation $\phi(t)$ represents the important parameter of optical source coherence; ω_{cw} and ω_{ccw} represent the angular frequency of sinusoidal phase modulation for CW and CCW waves respectively; M_{cw} and M_{ccw} represent the modulation index for CW and CCW waves respectively. Based on the theory of light wave field superposition, the intensities of two counter-propagating light waves I_{cw} and I_{ccw} are given by Equation (7) as

$$I_{cw} = A^2 I_0 \frac{1}{1 - Q^2} \cdot \frac{1 - Q_0^2}{1 + Q_0^2 - 2Q_0 cos(\varphi_{cw})}$$

$$I_{ccw} = A^2 I_0 \frac{1}{1 - Q^2} \cdot \frac{1 - Q_0^2}{1 + Q_0^2 - 2Q_0 cos(\varphi_{ccw})}$$
(8)

where *A* is the transmission coefficient of light propagating in FRR; φ_{cw} and φ_{ccw} are the phase shifts in the CW and CCW directions caused by the Kerr effect and phase modulation. The operating points for two light waves are always fixed at the resonance. Thus, considering the resonance behavior of Equation (8)

$$\varphi_{cw} = \varphi_{ccw} = 2m\pi \tag{9}$$

is obtained. Where *m* is an integer. Therefore, combining with Equation (6), the angular velocity drift Ω_{bias} induced by the Kerr effect is derived as

$$\Omega_{\text{bias}} = \frac{c\lambda_0\Lambda}{2\pi LD} \cdot \left(\frac{u_{cw}}{1 + \frac{1}{2}Q_0[M_{cw}\omega_{cw}\tau/(1 - Q_0)]^2} - \frac{u_{ccw}}{1 + \frac{1}{2}Q_0[M_{ccw}\omega_{ccw}\tau/(1 - Q_0)]^2}\right), \quad (10)$$
where
$$\begin{cases}
L = \frac{2\pi Zn_2}{\lambda_0A_{eff}} \cdot \frac{C_{cross}^2}{1 - Q^2} \cdot \frac{1 + Q_0}{(1 - Q_0)} \cdot \frac{1 - e^{-a_f L}}{a_f} \cdot e^{-a_c}I_0 \\
Q_0 = Q \cdot e^{-2\pi\Delta ft}
\end{cases}$$

and Δf is the linewidth of the light source. In addition, since we employ the same sinusoidal modulation scheme for both light waves, Equation (10) can be transformed into

$$\Omega_{\text{bias}} = \frac{c\lambda_0\Lambda}{2\pi LD} \cdot \left(\frac{u_{cw} - u_{ccw}}{1 + \frac{1}{2}Q_0[M_{cw}\omega_{cw}\tau/(1 - Q_0)]^2}\right).$$
(11)

With numerical evaluation, a plot of Ω_{bias} is shown in Figure 5 as a function of the ring resonator length *L*. Figure 5a shows that for the conventional RFOG with a single single-mode fiber (SMF) ring resonator [28] and Figure 5b for the PDHC-RFOG, respectively. The difference between the power-splitting coefficients $|u_{cw}-u_{ccw}|$ is 0.01%, and other parameters are listed in Tables 1 and 2.



Figure 5. Angular velocity drift Ω_{bias} induced by the Kerr effect as a function of the ring resonator length *L* in the case of different light source linewidths. (a) Drift of the conventional RFOG with a sing SMF ring resonator. (b) Drift of the PDHC-RFOG.

Table 2. Parameters for the RFOG with a single SMF ring resonator and the PDHC-RFOG.

Parameter	SMF-RFOG	PDHC-RFOG
n_0	1.45	1
f_m	50 kHz	50 kHz
М	2.405	2.405
α_c	0.5 dB	1 dB
α_f	0.2 dB/km **	20 dB/km *
Ž	259.81 **	380.53 *
n_2	$2.6 imes 10^{-20}~{ m m}^2/{ m W}$ **	$2.89 imes 10^{-22}~{ m m}^2/{ m W}$ *
$A_{e\!f\!f}$	$8.5 imes 10^{-11} \text{ m}^2 **$	$6.3 \times 10^{-11} \text{ m}^2 \text{ *}$

* Values provided by NKT Photonics. ** Values provided by Corning Incorporated.

Figure 5a illustrates that under the condition of constant linewidth of the light source, the drift Ω_{bias} induced by the Kerr effect decreases significantly as the ring resonator length *L* increases, while the system with a wider linewidth of the light source is less affected by the Kerr effect and can achieve lower drift Ω_{bias} with shorter fiber lengths. This is attributed to the change of the intensity in the FRR according to it fineness, which is closely dependent on its length and the light source linewidth. Hence, the Kerr effect noise has become the key performance barrier in the development of the RFOGs that require a shorter FRR length and narrower linewidth light source. Fortunately, this crucial problem can be settled by the PDHC-RFOG system. As shown in Figure 5b, the drift Ω_{bias} induced by the Kerr effect in the PDHC-RFOG is generally lower than that shown in Figure 5a.

Figure 6 compares the drift Ω_{bias} induced by the Kerr effect in a single SMF ring resonator and double HCPCF ring resonator, when the light source has a fixed linewidth of 50 kHz. The figure shows that the drift Ω_{bias} of the PDHC-RFOG is better than the conventional RFOG with a single SMF ring resonator. The reason for this conclusion is that (1), compared to the SMF, the HCPCF fiber has a smaller nonlinear refractive index coefficient, n_2 , and (2) the structure of the parallel-double ring resonator allows the CW and CCW light waves to propagate in the separate ring resonator, further reducing the optical power density in the ring resonator, decreasing the difference in optical intensity due to transmission, coupling, and environmental factors. As a result, the drift Ω_{bias} of the PDHC-RFOG is reduced by about three orders of magnitude compared to the conventional RFOG with a sing SMF ring resonator. The drift Ω_{bias} of the PDHC-RFOG is 8.54×10^{-7} rad/s when *L* is 10 m, indicating that the Kerr-effect-induced drift is smaller than the shot-noise-limited theoretical sensitivity.





5. Conclusions

For reducing many of the parasitic noises limiting the bias performance of the conventional RFOG with a single ring resonator, we propose a novel system structure of RFOG using a parallel-double HCPCF ring resonator, namely, the PDHC-RFOG, which can effectively solve various problems caused by the parasitic noises. The PDHC-RFOG can essentially eliminate the influence of the Rayleigh backscattering noise and theoretically reduce the Kerr-effect-induced drift by three orders of magnitude. In addition, we adopt the double closed loop and reciprocal modulation–demodulation technique, which not only overcomes the problem caused by the length mismatch between rings but also suppresses the second harmonic distortion induced by RAM noise and laser frequency noise. The sensitivity of the PDHC-RFOG can approach the shot-noise-limited theoretical value. Based on the preceding analysis, it is evident that the PDHC-RFOG can achieve high stability and high precision angular velocity measurement.

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