



Article Fault-Tolerant Control Strategy of a Wind Energy Conversion System Considering Multiple Fault Reconstruction

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Abstract: A robust sliding mode observer (SMO) is proposed to achieve multiple fault reconstruction for a wind energy conversion system (WECS) with simultaneous actuator and sensor faults. Firstly, the state equation of the WECS is established. The orthogonal transformation matrix and a post filter are introduced, and a new augmented system is constructed; then, the sensor fault is converted into an actuator fault to diagnose. The fault information is collected by the equivalent output control, and the simultaneous reconstruction algorithm of the sensor fault and the actuator fault is given. Through compensation control, the reliable control input of the WECS is guaranteed, and the function of active fault tolerant control for multiple faults is achieved. Simulation experiments show that the proposed method can accurately reconstruct the actuator and sensor faults, and maximum wind energy capture can be achieved by active fault-tolerant control.

Keywords: wind energy conversion system; sensor fault; actuator fault; multiple fault reconstruction; fault tolerant control

1. Introduction

Wind power generation is the most mature, largest and most promising power generation method for new energy [1]. Wind power plants are generally located in complex terrain and bad climates. The WECS, an important part of the wind power generation system, has a high fault rate, especially in its sensor and actuator, which seriously affect the performance of the system, even causing system paralysis, and cause immeasurable losses [2]. Therefore, realizing the fault tolerant control of WECS is an important guarantee for improving the operation reliability of wind power generation systems.

Fault-tolerant control aims to ensure a stable and safe operation of the controlled system under the requirements of a certain performance index when it fails [3]. Passive fault-tolerant control strategy cannot control the unknown faults of the system, while active fault tolerance can redesign the controller by using a priori knowledge of all kinds of faults [4]. Active fault-tolerant control is applied to various practical projects. An active fault-tolerant tracking controller is proposed for vehicle dynamics systems in [5], and the uncertain dynamic model of the vehicle is established. the vehicle dynamic states with sensor faults are estimated by the description observed and a fault-tolerant tracking controller based on linear matrix inequalities is designed. A robust H_{∞} output-feedback control strategy applied to the path following an autonomous ground vehicle is presented in [6], and a robust H_{∞} static output-feedback controller based on mixed genetic algorithms/linear matrix inequality is used to realize the path-following without the information of the lateral velocity. Active fault-tolerant control has also been widely studied and applied in WECS. An adaptive fault observer is designed for WECS, which detects the faults of the transmission part of the conversion system, and an active fault tolerant controller is designed to ensure the reliable and stable operation of the system [7]. An active tolerant

controller based on $H\infty$ is designed to obtain the high performance in extreme wind speed profile, and a sliding mode fault detection and isolation system is applied to reconstruct and compensate a wind turbine's sensor faults in [8].

The T-S fuzzy model proposed by Takagi T etc. is an effective method to solve the problem of nonlinear system control [9]. A proportional integral observer based on T-S fuzzy control is designed for the fault estimation of the actuator and sensor in [10]. Through the Lyapunov stability theory and L_2 performance analysis, the sufficient design conditions for simultaneous estimation of two kinds of faults are given, and the proposed conditions are solved under linear matrix inequalities constraints and the proportional integral observer gains are calculated. Aiming at the potential faults of the system, fuzzy modeling and the recognition method are used to automatically correct the potential faults of the system based on online diagnosis information, so as to ensure that the motor torque can be reliably adjusted when the actuator fails [11]. A multi-observer scheme based on Calman filter and maximum shift strategy is presented for the additive and multiplicative faults of sensors, and the fault types are diagnosed by generating new residual values in [12].

Sliding mode control can effectively deal with the uncertainty and interference of complex dynamic systems and adjust the dynamic behavior of the system by selecting sliding function [13] so that the nonlinear system has strong robustness, and this is widely applied in engineering practice, such as CNC machine tools [14], four-rotor helicopters [15], wind turbines [16] and so on. In [17], a two-order sliding mode is designed for the parameter variable and external disturbance of the wind energy system based on doubly-fed induction generators DFIG, and the variable structure control based on the two-order derivative of the a properly chosen sliding manifold is designed to further improve the performance of the system. In [18], a class of MIMO (multiple-input multiple-output) nonlinear system's derivative and integral terminal sliding mode control is proposed to realize the limited convergence time of the more general high order MIMO system and avoid the singular problems in the design of the controller. Fault detection and isolation for induction generators based on a sliding mode observer is proposed in [19] by means of maximum power point tracking, and the induction generator closed loop via indirect rotor flux-oriented control is described. The WECS has the characteristics of being nonlinear, having strong coupling, being multivariable and so on, and is seriously disturbed by external interference. Through the sliding mode control, the WECS has strong robustness and excellent insensitivity to faults and external interference.

The WECS is usually composed of a wind turbine, transmission link, generator, Alternating Current-Direct Current (AC-DC) converter and so on; there has been much research on fault diagnosis and fault-tolerant control of different links in the WECS. A Wound Rotor Induction Generator (WRIG)-based wind energy system associated with hybrid AC/DC microgrid (DCM) is introduced in [20], which runs under an islanded and utility-tied mode. The system operation and power transfer under the islanded and utility-tied mode are realized at all possible conditions, and the effectiveness of the system in the process of load changes is proved by the dynamic experimental results of various modes. For MPP (maximum power point) and no-MPP tracking operations, two different methods, the difference between the actual and simulated output power of the wind system and the predetermined angular shaft speed ratio are adopted respectively as indictors to detect the lubricant system faults in [21]. According to different operation modes, different diagnostic criteria are used to diagnose faults; although it is helpful for better detection of faults, the process design is rather complicated. An active fault-tolerant fuzzy control strategy based on multi observer switching control is proposed in [22], the T-S fuzzy modeling of WECS is carried out and sufficient conditions for robust stabilization of system are given by deriving Taylor series and linear matrix inequalities. Multi-observer switching control leads to an increase in the number of constraints, which is limited in engineering applications, and the actuator fault is not considered in it. A fault diagnosis method for the lubrication system and the converter in the wind energy system is presented in [23], and the tolerance of the supervisory controller towards a lubrication system and converter fault in the wind system is explained to satisfy the demanded power with maximum utilization. Compared with proposed robust formulation, this

paper has been optimized to better solve the uncertainty of the WECS. With the increase of the scale of WECS, the number of actuators and sensors is increasing, and the probability of simultaneous faults is increasing. Previous works focused on fault diagnosis and fault tolerant control of a single actuator or sensor. Therefore, it is more practical to study the multi-fault diagnosis of WECS.

In this paper, the state equation of the WECS is established in the case of the simultaneous faults of the actuator and the sensor in the WECS. Based on the SMO, a new augmented system is constructed by introducing the orthogonal transformation matrix and constructing the post filter. The sensor fault is converted into the actuator fault to diagnose, and a robust adaptive fault reconstruction observer is designed to achieve multiple fault reconstruction. A robust fault-tolerant controller for WECS is designed to guarantee the reliable control input of the WECS and to capture the maximum wind energy. The simulation results are given to verify the superiority of the proposed method.

2. Mathematical Modeling of WECS Based on DFIG

Doubly-fed induction generators (DFIG) are widely used in wind power systems due to their stability, robustness and relatively low cost. As a result, the doubly fed wind power generation system is taken as an example [24]—WECS based on DFIG—as shown in Figure 1. The wind power is converted to mechanical energy by a wind turbine, and the DFIG is driven by the drive system to generate electric energy and then transmitted to the power grid through an AC-DC converter.



Figure 1. Wind energy conversion system (WECS) based on doubly-fed induction generators (DFIG).

Assuming that the wind turbine is an ideal state, according to Baez theory, the mechanical power of the wind turbine is as follows [25]:

$$P_{wt} = 0.5\pi R^2 \rho v^3 C_p(\lambda, \beta) \tag{1}$$

The wind torque produced by the wind turbine is as follows [26]:

$$\Gamma_{wt} = \frac{P_{wt}}{\Omega_l} = 0.5\pi\rho v^2 R^3 C_{\Gamma}(\lambda,\beta)$$
⁽²⁾

where ρ is the air density, v is the wind speed, R is the blade length, $C_p(\lambda,\beta)$ is the conversion coefficient of wind energy. λ is the ratio of blade tip velocity to wind speed; that is, $\lambda = \Omega_l(R/v)$, Ω_l is the angular velocity of wind turbine rotor(low speed shaft), and $C_{\Gamma}(\lambda,\beta)$ is the torque coefficient, satisfying $C_{\Gamma}(\lambda,\beta) = C_P(\lambda,\beta)/\lambda$.

In order to simplify the modeling, the rigid model is adopted in the connection between the high-speed axis and the low speed axis [27]; the equation of motion of the transmission system Equations (3) and (4) is as follows:

$$J_h \frac{d\Omega_h}{d_t} = \frac{\eta}{i_0} \Gamma_{wt} - \Gamma_G \tag{3}$$

$$J_t \frac{d\Omega_l}{d_t} = \Gamma_{wt} - \frac{i_0}{\eta} \Gamma_G \tag{4}$$

where Ω_h is the rotor speed (high speed shaft) of generator, that is, $\Omega_h = i_o \times \Omega_l$, i_o is the gear transmission ratio, Γ_G is the electromagnetic torque of generator, η is the transmission efficiency, J_h is the high speed shaft inertia, J_t is the low speed shaft inertia.

Some dynamic characteristics of the model are neglected and, considering that the electromagnetic time constant is much smaller than the mechanical time constant, the dynamic process of electromagnetic response of generator is ignored when modeling the system [28]. The state equation of WECS is modeled as follows:

$$\begin{cases} \dot{x}_p(t) = A_p x_p(t) + B_p u(t) \\ y_p(t) = C_p x_p(t) \end{cases}$$
(5)

where $x_p(t) = \begin{bmatrix} \Omega_h & \Gamma_G \end{bmatrix}^T$, $u(t) = \Gamma_{ref}^*$, $A_p = \begin{bmatrix} \frac{\Gamma_{wt}(i_0\Omega_h,v)}{i_0\cdot J_t\cdot\Omega_h} & -\frac{1}{J_t} \\ 0 & -\frac{1}{T_G} \end{bmatrix}$, $B_p = \begin{bmatrix} 0 \\ \frac{1}{T_G} \end{bmatrix}$, $C_p = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$;

 $x_p(t)$ is the state vector, u(t) is the input vector, $y_p(t)$ is the output vector, Γ_{ref}^* is the reference value of the electromagnetic torque of the generator, and T_G is the electromagnetic time constant [29].

3. Sensor Fault Conversion

The faults of WECS include system faults, actuator faults, sensor faults and so on. Actuator faults usually occur due to wear and destruction of gears in gear boxes, and wear and deformation of bearing surfaces. Sensors usually fail due to their aging damage and uncorrected long-term use. Considering the faults of the WECS, the system (5) can be described as

$$\begin{cases} \dot{x}_{p}(t) = A_{p}x_{p}(t) + B_{p}u(t) + F_{p}f_{a}(t) \\ y_{p}(t) = C_{p}x_{p}(t) + G_{p}f_{s}(t) \end{cases}$$
(6)

where $A_p \in \mathbb{R}^{n \times n}$, $B_p \in \mathbb{R}^{n \times m}$, $C_p \in \mathbb{R}^{p \times n}$, $F_p \in \mathbb{R}^{n \times q}$, $G_p \in \mathbb{R}^{p \times h}$, $q \le m , <math>C_p$ and F_p are full rank matrixes, $f_a(t)$ is the actuator fault, $f_s(t)$ is the sensor fault, F_p is the actuator fault distribution matrix, and G_p is the sensor fault distribution matrix.

Assumption 1. $f_a(t)$ and $f_s(t)$ are norm bounded.

Assumption 2. System Equation (6) satisfies $rank(C_pF_p) = rank(F_p) = q$.

Assumption 3.
$$rank \begin{bmatrix} sI - A_p & F_p \\ C_p & 0 \end{bmatrix} = n + q.$$

Based on system Equation (6), the orthogonal matrix is introduced [30] and can be obtained as follows:

$$T_R := \begin{cases} y_1 = C_{p_1} x_p(t) \\ y_2 = C_{p_2} x_p(t) + G_1 f_s(t) \end{cases}$$
(7)

State variable Equation (8) is defined as follows:

$$\dot{z}_p(t) = -A_f z_p(t) + A_f y_2(t) = -A_f z_p(t) + A_f C_{p_2} x_p(t) + A_f G_1 f_s(t)$$
(8)

where $z_p \in \mathbb{R}^h$ is the state variable of y_2 by the filter, $G_1 \in \mathbb{R}^{h \times h}$ is a nonsingular matrix, $A_f \in \mathbb{R}^{h \times h}$ is designed to filter the matrix, and Equations (7) and (8) are brought into the Equation (6), and the new system equation is obtained:

$$\begin{cases} \begin{bmatrix} \dot{x}_{p}(t) \\ \dot{z}_{p}(t) \end{bmatrix} = \begin{bmatrix} A_{p} & 0 \\ A_{f}C_{p_{2}} & -A_{f} \end{bmatrix} \begin{bmatrix} x_{p}(t) \\ z_{p}(t) \end{bmatrix} + \begin{bmatrix} B_{p} \\ 0 \end{bmatrix} u(t) + \begin{bmatrix} F_{p} & 0 \\ 0 & A_{f}G_{1} \end{bmatrix} \begin{bmatrix} f_{a}(t) \\ f_{s}(t) \end{bmatrix} \\ \begin{bmatrix} y_{1}(t) \\ z_{p}(t) \end{bmatrix} = \begin{bmatrix} C_{p_{1}} & 0 \\ 0 & I_{h} \end{bmatrix} \begin{bmatrix} x_{p}(t) \\ z_{p}(t) \end{bmatrix}$$
(9)

Equation (9) is simplified and the new state variables and the system matrices are defined as follows:

$$\dot{x}(t) = \begin{bmatrix} \dot{x}_p(t) \\ \dot{z}_p(t) \end{bmatrix}, x(t) = \begin{bmatrix} x_p(t) \\ z_p(t) \end{bmatrix}, A = \begin{bmatrix} A_p & 0 \\ A_f C_{p_2} & -A_f \end{bmatrix}, B = \begin{bmatrix} B_p \\ 0 \end{bmatrix}, F = \begin{bmatrix} F_p & 0 \\ 0 & A_f G_1 \end{bmatrix},$$

$$f_{a,s}(t) = \begin{bmatrix} f_a(t) \\ f_s(t) \end{bmatrix}, C = \begin{bmatrix} C_{p_1} & 0 \\ 0 & I_h \end{bmatrix}.$$
Equation (9) can be transformed into:

uation (9) can be transformed into:

$$\begin{cases} \dot{x}(t) = Ax(t) + Bu(t) + Ff_{a,s}(t) \\ y(t) = Cx(t) \end{cases}$$
(10)

Equation (10) is a system with an actuator fault $f_a(t)$ and sensor fault $f_s(t)$, which is equivalent to actuator faults in Equation (6).

There is a nonsingular transformation matrix $T_p \in R^{(n+h) \times (n+h)}$ [31], $T_p x = \begin{vmatrix} x_1 \\ x_2 \end{vmatrix}$, which places the original system Equation (10) under the action of T_p , to obtain Equation (11):

$$\begin{cases} \dot{x}_1(t) = A_1 x_1(t) + A_2 x_2(t) + B_1 u(t) \\ \dot{x}_2(t) = A_3 x_1(t) + A_4 x_2(t) + B_2 u(t) + F_2 f_{a,s}(t) \\ y(t) = x_2(t) \end{cases}$$
(11)

where $x_1 \in R^{n+h-p}$, $x_2 \in R^p$, $T_pAT_p^{-1} = \begin{bmatrix} A_1 & A_2 \\ A_3 & A_4 \end{bmatrix}$, $A_1 \in R^{(n+h-p)\times(n+h-p)}$, $A_2 \in R^{(n+h-p)\times p}$, $A_3 \in R^{p\times(n+h-p)}$, $A_4 \in R^{p\times p}$; $T_pB = \begin{bmatrix} B_1 \\ B_2 \end{bmatrix}$, $B_1 \in R^{(n+h-p)\times(n+h-p)}$, $B_2 \in R^{p\times(n+h-p)}$; $T_pF = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$ $\begin{bmatrix} 0 \\ F_2 \end{bmatrix}$, $F_2 \in R^{p \times (q+h)}$; $CT = \begin{bmatrix} 0 & I_p \end{bmatrix}$. For the proof, refer to [16].

4. Design and Reconstruction of Sliding Mode Fault Observer (SMFRO)

4.1. Design of Robust Adaptive Sliding Mode Fault Observer

For the system Equation (10), the sliding mode state observer is designed as follows:

$$\begin{cases} \dot{\hat{x}}(t) = A\hat{x}(t) + Bu(t) - L_l(C\hat{x}(t) - y(t)) + L_n \nu_\rho \\ \hat{y}(t) = C\hat{x}(t) \end{cases}$$
(12)

where L_l is the undetermined gain matrix, L_n is the nonlinear feedback matrix, and ν_{ρ} is the sliding mode strategy to be designed. The state estimation error of the observer is defined as $e(t) = \hat{x}(t) - x(t)$, the output estimation error is $e_y(t) = \hat{y}(t) - y(t)$, and the deviation system equation $\dot{e}(t) = (A - A)$ $L_lC)e(t) + L_n\nu_{\rho} - Ff_{a,s}(t)$ is obtained by Equations (10) and (12).

According to the system Equation (11), a sliding mode fault reconstruction observer is proposed:

$$\begin{cases} \dot{x}_1(t) = A_1 \hat{x}_1(t) + A_2 \hat{x}_2(t) + B_1 u(t) - A_2 e_y(t) \\ \dot{y}_2(t) = A_3 \hat{x}_1(t) + A_4 \hat{x}_2(t) + B_2 u(t) - (A_4 - A_4^s) e_y(t) + \nu_\rho \end{cases}$$
(13)

where $A_4^s \in \mathbb{R}^p$ is the arbitrary stable matrix, the robust SMO is undetermined gain matrix L_l and the nonlinear feedback matrix L_n is, respectively, $L_l = T^{-1} \begin{bmatrix} A_2 \\ A_4 - A_4^s \end{bmatrix}$, $L_n = T^{-1} \begin{bmatrix} 0 \\ I_p \end{bmatrix}$.

It is assumed that the deviation between the system (10) and the SMO (12) is always bounded: $||e|| \le \varepsilon$, ε is the normal number, and the following sliding mode strategy is proposed for the observer Equation (12):

$$\nu_{\rho} = \begin{cases} -\rho \|P_2\| \frac{P_2 C L_n e_y}{\|P_2 C L_n e_y\|} & e_y \neq 0\\ 0 & e_y = 0 \end{cases}$$
(14)

where $P_2 \in R^p$ is the Lyapunov matrix of A_4^s , ρ is the correction coefficient and $0 < \rho \leq 1$.

The state estimation error is $e_1(t) = \hat{x}_1(t) - x_1(t)$, and the output deviation is $e_y(t) = \hat{x}_2(t) - x_2(t)$. The state and output equations of the system are Equations (15) and (16):

$$\dot{e}_1 = A_1 e_1(t)$$
 (15)

$$\dot{e}_y = A_3 e_1(t) + A_4^s e_y(t) + \nu_\rho - F_2 f_{a,s}(t)$$
(16)

According to [32], the system Equations (15) and (16) are asymptotically stable. When $(e_1, e_y) \rightarrow 0$, the SMO is asymptotically stable.

4.2. Robust Multiple Fault Reconstruction

The discontinuous switch v_{ρ} , i.e., sliding mode strategy Equation (14), makes the observer deviation equation produce a sliding mode motion to ensure the robustness of the system to the external disturbance, but it will inevitably produce chattering and bring high frequency interference to the system. When the sliding mode motion reaches $\dot{e}_y = e_y = 0$; i.e., $0 = A_3 e_1(t) + v_{eq} - F_2 f_{a,s}(t)$, where v_{eq} is the equivalent output control. Because A_1 has a stable pole, it makes $\lim_{t\to\infty} e_1 = 0$, then $v_{eq} \to F_2 f_{a,s}(t)$. In order to reduce the chattering phenomenon, the continuous function approximation method is used to approximate the sliding mode and Equation (17) is obtained [33]:

$$\nu_{eq} = -\rho \|P_2\| \frac{P_2 C L_n e_y}{\|P_2 C L_n e_y\| + \sigma}$$
(17)

where σ is the normal number with less accuracy. The equivalent output control Equation (17) makes the reconfiguration system robust, and the robust fault reconstruction expression is as follows:

$$\hat{f}_{a,s}(t) = -\rho \|P_2\| (F_2^T F_2)^{-1} F^T \frac{P_2 C L_n e_y}{\|P_2 C L_n e_y\| + \sigma}$$
(18)

The robust fault reconstruction expressions for the actuator and the sensor are Equations (19) and (20):

$$\hat{f}_{a}(t) = \begin{bmatrix} I_{q} & 0 \end{bmatrix} \hat{f}_{a,s}(t)$$
 (19)

$$\hat{f}_{s}(t) = \begin{bmatrix} 0 & I_{h} \end{bmatrix} \hat{f}_{a,s}(t)$$
 (20)

5. Design and Reconstruction of Sliding Mode Fault Observer (SMFRO)

A sliding surface expression Equation (21) is designed for WECS:

$$\sigma = a_1 J_t \Omega + a_2 J_t \Gamma_G - J_t \dot{\Omega} \tag{21}$$

where a_1 is the time constant of the convergent speed of sliding mode control, that is, $\lambda_{opt} : \Omega = a_1 \Omega_{opt} + a_2 \Gamma_{opt} = 0$, then $a_2 = -a_1 \frac{\Omega_{opt}}{\Gamma_{opt}}$. A stable sliding surface can be obtained by the Equation (21).

The sliding mode control law consists of the equivalent control input U_{eq} and the switch part U_n two parts are as follow [29]:

$$U = U_{eq} + U_n \tag{22}$$

$$U_{eq} = \Gamma_G - \frac{T_G}{1 + a_2 J_t} (a_1 J_t \Omega_h + a_2 J_t \Gamma_G) (a_1 - A(\lambda, v))$$

$$\tag{23}$$

$$U_n = -\alpha \operatorname{sgn}_h(\sigma) \tag{24}$$

where $A(\lambda, v) = \frac{KvR^2}{i} \cdot \frac{\lambda \cdot C'_p(\lambda) - C_p(\lambda)}{\lambda^2}$, $K = 0.5\pi\rho R^2$, $C'_p(\lambda)$ is the derivative of the power coefficient λ , and $\operatorname{sgn}_h(\sigma)$ is the hysteresis function of the bandwidth of h.

The multi fault reconstruction block diagram of the WECS is shown in Figure 2.



Figure 2. Multi fault reconstruction control block diagram of WECS.

The fault output of the sliding mode controller is as follows:

$$U_1 = U + f_{a,s}(t)$$
 (25)

The output of the sliding mode fault reconstruction observer is as follows:

$$U_2 = U + \hat{f}_{a,s}(t)$$
 (26)

The control input of the WECS is obtained by the Equations (25) and (26).

$$\Gamma_{ref}^* = U + U_1 - U_2 \tag{27}$$

6. Experiments and Results

A three-lobe horizontal axis doubly fed wind energy conversion system with low power (6 kw), high speed and fixed pitch is used [30]. The simulation is carried out by Matlab software. The simulation parameters are shown in Table 1 as follows:

Under the rated wind speed [27], the fixed pitch control is adopted, i.e., $\beta = 0^{\circ}$, and the wind energy conversion coefficient is determined by Equation (28):

$$C_p(\lambda,\beta) = C_p(\lambda) = -4.54 \times 10^{-7} \lambda^7 + 1.3027 \times 10^{-6} \lambda^6 -6.5416 \times 10^{-5} \lambda^5 - 9.7477 \times 10^{-4} \lambda^4 +0.0081 \lambda^3 - 0.0013 \lambda^2 + 0.0061 \lambda$$
(28)

In the case of a certain wind speed, when the blade tip speed ratio is $\lambda = 7$, the maximum value is 0.476: the best tip speed ratio.

Parameter Symbols	Parameter Names	Parameter Values
V_s	Rated voltage	220 V
w_s	Rated speed	100 πrad/s
Γ _{Gmax}	Rated electromagnetic torque	40 N·m
ρ	Air density	1.25 kg/m ³
η	Transmission efficiency	95%
i ₀	Transmission speed ratio	6.25
R	Blade length	2.5 m
J_h	Moment of inertia of high speed axis	0.092 kg⋅m ²
Jt	Moment of inertia of low speed axis	$5.49 \text{ kg} \cdot \text{m}^2$
T _G	Electromagnetic time constant	0.02

Table 1. Simulation Parameters.

The wind speed of the WECS is shown in Figure 3.





When there are no faults in the WECS, the reference value of the electromagnetic torque is Γ_{ref}^* , and the wind energy conversion coefficient is C_p , as shown in Figures 4 and 5.



Figure 4. Reference value of electromagnetic torque without fault Γ_{ref}^* .



Figure 5. Wind energy conversion coefficient without fault.

When the wind energy conversion system fails, the expression of the actuator fault is as follows:

$$f_a = \begin{cases} 12\sin(\pi t), & 24 \le t \le 40\\ 0, & otherwise \end{cases}$$
(29)

The fault time is set to $24{\sim}40 \text{ s}$.

The expression of sensor fault is

$$f_s = \begin{cases} 16\sin(0.6\pi t), & 32 \le t \le 48\\ 0, & otherwise \end{cases}$$
(30)

The fault time is set to 32~48 s.

The proposed fault reconstruction algorithm is used to reconstruct the actuator fault. The reconfiguration of actuator fault is shown in Figure 6 and the reconfiguration of sensor fault is shown in Figure 7. In the range of error allowed, the online tracking and reconfiguration of the fault can be realized.

When the actuator fault Equation (30) and the sensor fault occurs, the electromagnetic torque reference value Γ_{ref}^* and the wind energy conversion coefficient C_p are shown in Figures 8 and 9, respectively.



Figure 6. Actuator fault and reconfiguration.



Figure 7. Sensor fault and reconfiguration.



Figure 8. The reference value of electromagnetic torque Γ_{ref}^* at the time of fault.



Figure 9. The wind energy conversion coefficient C_p at the time of fault.

When the fault occurs, the wind energy conversion coefficient varies greatly, which affects the capture of wind energy. After adopting an active fault-tolerant controller, the capture efficiency is significantly improved and the purpose of optimization is achieved. The reference value of electromagnetic torque Γ_{ref}^* and wind energy conversion coefficient C_p after fault-tolerant control are shown in Figures 10 and 11. Compared with Figures 8 and 9 and Figures 10 and 11, the expected value of the wind energy conversion coefficient is reached and the maximum wind energy capture is achieved after active fault tolerance control.



Figure 10. The reference value of electromagnetic torque Γ_{ref}^* after fault-tolerant control.



Figure 11. The wind energy conversion coefficient C_p after fault-tolerant control.

7. Conclusions

In this paper, the problem of system fault reconstruction and fault-tolerant control is discussed when multiple faults occur in the WECS. Firstly, the state equation of the WECS is established. Based on the SMO, a new augmented system is constructed by orthogonal transformation matrix and post filter, and the sensor fault is converted into the actuator fault to diagnose. A robust adaptive SMO is designed, and an optimized sliding mode strategy is given to ensure the stability of the state estimation of the system. The fault information is collected in real time by the equivalent output control, and the simultaneous reconstruction algorithm of the sensor fault and the actuator fault is given. Finally, a robust fault-tolerant controller is designed to ensure reliable control input of the WECS. When sensor and actuator faults occur, fault-tolerant control can be carried out successfully to achieve maximum wind energy capture.

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