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Fault-Tolerant Control for Actuator Faults of Wind Energy Conversion System

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Abstract: The problem of robust fault-tolerant control for actuators of nonlinear systems with uncertain parameters is studied in this paper. Takagi–Sugeno (T-S) fuzzy model is used to describe the wind energy conversion system (WECS). Fuzzy dedicated observer (FDO) and fuzzy proportional integral observer (FPIO) are established to reconstruct the system state and actuator fault, respectively. Fuzzy Robust Scheduling Fault-Tolerant Controller (FRSFTC) is designed by parallel distributed compensation (PDC) method, so as to realize the purpose of active fault tolerance for actuator faults and ensure the robust stability of the system. The stability of the closed-loop system is proved by Taylor series, Lyapunov function, and Linear Matrix Inequalities (LMIs). Finally, the simulation results verify that the proposed method is feasible and effective applied to WECS with doubly fed induction generators (DFIG).

Keywords: WECS; T-S fuzzy; parameter uncertainty; actuator failure; fault-tolerant control

1. Introduction

With the development of science and technology, many high-level comprehensive complex nonlinear systems have been applied at a large scale. It is becoming increasingly important to have certain fault tolerance for the control system to ensure the safety and reliability of the control system. As an important green energy, wind energy development and use has become a focus in the global energy field [1–4]. The wind energy conversion system (WECS) is a typical large and complex nonlinear system with random and intermittent wind force. A WECS fault can be categorized into three types—transmission system fault, converter fault, and power grid fault—among which the transmission system, because of its structural characteristics, produces tooth surface wear, fatigue erosion, broken teeth and bearing resistance, planetary wheel cracking, and other problems [5,6]. Therefore, the transmission system has the highest probability of failure, which has the greatest impact on the WECS performance. To ensure the safe and efficient operation of wind turbines, fault-tolerant control (FTC) becomes a hotspot of research regarding WECS.

Currently, in the fault-tolerant control of WECS, the H ∞ infinity fault-tolerant control method for WECS was studied based on the random piece wise affine (PWA) model. The modeling and fault-tolerant control of WECS under random wind load were solved [7]. The research illustrates that this method has better fault tolerance for sensor and actuator gain faults in WECS; the angle estimation of the pitch system is carried out by Kalman filter, to detect the fault of the blade pitch system and achieve satisfactory detection results [8]. The results show that this method can improve the efficiency of wind energy capture; on the basis of the adaptive fault observer, a state feedback fault-tolerant controller was designed to ensure the good performance of the system in normal operation in case of failure [9]. The essence of T-S fuzzy model is to use if-then fuzzy inference rules to describe the nonlinear system. Each inference rule stands for the dynamics of the local regional linear model. Using the membership function to connect the local linear models to obtain the overall fuzzy linear model, and then achieve the purpose of system modeling. In recent years, T-S fuzzy model has been widely used in controller design and system performance analysis of uncertain nonlinear systems due to its advantages such as simple structure and strong approximation, which enables it to approximate almost any complex nonlinear system [10–15]. A reliable hybrid H∞/passive control problem for T-S fuzzy time-delay systems was studied based on the semi-Markov jump model (SMJM) [16]; T-S model was used to describe the offshore wind power system, and proposing an active fault-tolerant tracking control method for sensor faults [17]; T-S modeling of doubly fed wind power system was also carried out [18,19]. For the actuator or sensor faults of the system, fault-tolerant controllers were established based on fuzzy synovial observer and fuzzy observer, respectively. However, the robust stability of the system cannot be fully guaranteed because the uncertainty generated by the system were not taken into account in the literature.

In this study, T-S fuzzy model is used to describe the WECS, and the unmeasurable state variables and uncertainties in the system are taken into account. Based on the T-S fuzzy model, the system state is estimated effectively by fuzzy dedicated observer (FDO). Fuzzy proportional integral observer (FPIO) is designed for system actuator fault to realize accurate reconstruction of fault signal. According to the estimated fault information, a robust scheduling fault-tolerant controller is designed by using parallel distributed compensation (PDC) method to realize real-time compensation for actuator faults of uncertain nonlinear systems. Taylor series and Lyapunov stability theory are used to prove the necessary and sufficient conditions to keep the closed-loop stability of the system. Finally, the feedback gain matrices are obtained via addressing Linear Matrix Inequalities (LMIs). Simulation results further verify the proposed control strategy is reliability and effectiveness applied to WECS.

2. Problem Description

2.1. Design of TS Fuzzy Model

A series of fuzzy rules are established for the nonlinear system with uncertainty. Each rule represents one of the subsystems, so the T-S model structure of uncertain parameters fuzzy system Equation (1) is described as follows:

 R^i : If $z_1(t)$ is F_1^i and $z_2(t)$ is F_2^i ... and $z_k(t)$ is F_k^i , then

$$\dot{x}(t) = (A_i + \Delta A_i)x(t) + (B_i + \Delta B_i)u(t)$$

$$y(t) = C_i x(t) \qquad i = 1, 2 \dots r$$
(1)

where R^i is the *i*th fuzzy inference rule, $z(t) = [z_1(t) \ z_2(t) \dots z_k(t)]^T$ denotes the premise variables, F_j^i is the fuzzy sets; $i = 1, 2, \dots, r$ is the number of system fuzzy inference rules, $j = 1, 2, \dots, k$, $x(t) \in R^n$ is the state of the system, $u(t) \in R^m$ represents the control input, $y(t) \in R^p$ is the system output; $A_i \in R^{n \times n}$, $B_i \in R^{n \times m}$ and $C_i \in R^{p \times n}$ are the parameter matrices of the system, ΔA_i and ΔB_i are the uncertain real value matrices. It is assumed that the uncertainty norm of the system is bounded. The entire equation of state of the whole fuzzy T-S system Equation (2) can be obtained after inverse fuzzification:

$$\dot{x}(t) = \sum_{i=1}^{r} u_i(z(t)) [(A_i + \Delta A_i)x(t) + (B_i + \Delta B_i)u(t)]$$

$$y(t) = \sum_{i=1}^{r} u_i(z(t))C_ix(t)$$
(2)

where $h_i(z(t)) = \prod_{j=1}^k F_j^i(z_j(t)), u_i(z(t)) = \frac{h_i(z(t))}{\sum\limits_{i=1}^r h_i(z(t))}$. Hence, $F_j^i(z_j(t))$ represents the membership function of $z_j(t)$ on fuzzy sets F_{j}^i , $h_i(z(t))$ is the weight of rule *i*, $h_i(z(t))$ and $u_i(z(t))$ satisfies the following condition: $h_i(z(t)) \ge 0$, $\sum_{i=1}^{N} h_i(z(t)) > 0$, $i = 1, 2, ..., r, 0 < u_i(z(t)) < 1$, $\sum_{i=1}^{Nr} u_i(z(t)) = 1$. Considering the failure of the system actuator, the system model (2) is rewritten as follows:

$$\dot{x}(t) = \sum_{i=1}^{r} u_i(z(t))[(A_i + \Delta A_i)x(t) + (B_i + \Delta B_i)u(t) + D_id(t)]$$

$$y(t) = \sum_{i=1}^{r} u_i(z(t))C_ix(t)$$
(3)

where $D_i = B_i \overline{D}_i$, $\overline{D}_i \in R^{m \times q}$ represents the known actuator fault matrix, $d(t) \in R^{q \times 1}$ is the time-varying signal of actuator fault (q < n), the premise variable $z_1(t) z_2(t) \dots z_k(t)$ is measurable and fault independent.

In this paper, let

$$\sum_{i=1}^{r} u_i(z(t)) \Delta A_i = \Delta A = \begin{pmatrix} \Delta a_{11} & \dots & \Delta a_{1n} \\ \vdots & \ddots & \vdots \\ \Delta a_{n1} & \dots & \Delta a_{nn} \end{pmatrix}, \sum_{i=1}^{r} u_i(z(t)) \Delta B_i = \Delta B = \begin{pmatrix} \Delta b_{11} & \dots & \Delta b_{1m} \\ \vdots & \ddots & \vdots \\ \Delta b_{n1} & \dots & \Delta b_{nm} \end{pmatrix}$$

Furthermore, the system model (3) can be expressed as:

$$\dot{x}(t) = \sum_{i=1}^{r} u_i(z(t)) [(A_i + \Delta A)x(t) + (B_i + \Delta B)u(t) + D_i d(t)]$$
(4)

$$y(t) = \sum_{i=1}^{r} u_i(z(t))C_i x(t)$$
(5)

2.2. Uncertain Fuzzy Parameters

System uncertainty *l*th fuzzy rule is described by:

Rule *l*: If Δa_{11} is $N_{\Delta a_{11}}^l$ and \ldots and Δa_{nn} is $N_{\Delta a^{nn}}^l$ and Δb_{11} is $N_{\Delta b_{11}}^l$ and Δb_{nm} is $N_{\Delta b_{nm}}^l$, then $\Delta A = \Delta \widetilde{A}_l$, $\Delta B = \Delta B_1$.

The output of fuzzy uncertainty can be expressed as:

$$\Delta A = \sum_{l=1}^{s} h_l(\Delta A, \Delta B) \Delta \widetilde{A}_l, \Delta B = \sum_{l=1}^{s} h_l(\Delta A, \Delta B) \Delta \widetilde{B}_l$$
(6)

where $\sum_{l=1}^{s} h_l(\Delta A, \Delta B) = 1, h_l(\Delta A, \Delta B) \in [0 \ 1]$ for $\forall 1, h_l(\Delta A, \Delta B) = \frac{\varpi_l(\Delta a, \Delta b)}{\sum\limits_{l=1}^{s} \varpi_l(\Delta a, \Delta b)}$. $\varpi_l(\Delta a, \Delta b) = N_{\Delta a_{11}}^l(\Delta a_{11}) \times \ldots \times N_{\Delta a_{nn}}^l(\Delta a_{nn}) \times N_{\Delta b_{11}}^l(\Delta b_{11}) \times \ldots \times N_{\Delta a_{nm}}^l(\Delta a_{nm})$ (7)

 $s = 2^{c}$ represents the number of fuzzy rules; the scalar *c* stands for the number of uncertain elements in ΔA and ΔB , ΔA_l and ΔB_l are given by:

$$\Delta \widetilde{A}_{l} = \begin{pmatrix} \Delta a_{11}^{\max ormin} & \dots & \Delta a_{1n}^{\max ormin} \\ \vdots & \ddots & \vdots \\ \Delta a_{n1}^{\max ormin} & \dots & \Delta a_{nn}^{\max ormin} \end{pmatrix}, \ \Delta \widetilde{B}_{l} = \begin{pmatrix} \Delta b_{11}^{\max ormin} & \dots & \Delta b_{1m}^{\max ormin} \\ \vdots & \ddots & \vdots \\ \Delta b_{n1}^{\max ormin} & \dots & \Delta b_{nm}^{\max ormin} \end{pmatrix}$$

Define fuzzy weights: $\sum_{i=1}^{r} u_i(z(t)) = \sum_{l=1}^{s} h_l(\Delta A, \Delta B) = \sum_{i=1}^{r} \sum_{l=1}^{s} u_i(z(t))h_l(\Delta A, \Delta B) = 1.$ In the following, Let's write $h_l(\Delta A, \Delta B)$ and $u_i(z(t))$ as h_l and u_i respectively for the sake of

In the following, Let's write $h_l(\Delta A, \Delta B)$ and $u_i(z(t))$ as h_l and u_i respectively for the sake of simplifying writing, from Equations (4)–(6), the system model (3) becomes:

$$\dot{x}(t) = \sum_{i=1}^{r} \sum_{l=1}^{s} u_i h_l \Big[(A_i + \Delta \widetilde{A}_l) x(t) + (B_i + \Delta \widetilde{B}_l) u(t) + D_i d(t) \Big]$$
(8)

$$y(t) = \sum_{i=1}^{r} u_i C_i x(t)$$
 (9)

2.3. Nonlinear FPIO

For T-S fuzzy system actuator fault, FPIO is designed based on T-S fuzzy model. Where the *i*th fuzzy rule R^i is: If $z_1(t)$ is F_1^i and $z_2(t)$ is $F_2^i \dots$ and $z_{\psi}(t)$ is F_{ψ}^i , then

$$\dot{\hat{x}}_{u}(t) = A_{i}\hat{x}_{u}(t) + B_{i}u(t) + K_{i}(y(t) - \hat{y}_{u}(t)) + D_{i}\hat{d}(t)
\hat{y}_{u}(t) = C_{i}\hat{x}_{u}(t) \qquad i = 1, 2, \dots, p$$
(10)

The estimated value of actuator fault time-varying signal d(t) can be expressed by:

$$\hat{d}(t) = G_i(y(t) - \hat{y}_u(t)) = G_i \tilde{y}(t)$$
(11)

where $\dot{x}_u(t)$ represents the estimated state by FPIO, K_i are observation error matrices, G_i are their integral gains to be designed, y(t) is the output vector, $\hat{y}_u(t)$ is the final output of the FPIO, $\tilde{y}(t)$ is the output estimation error. After defuzzification, the final output of FPIO Equation (12) is described as follows:

$$\dot{\hat{x}}_{u}(t) = \sum_{i=1}^{r} u_{i}[A_{i}\hat{x}_{u}(t) + B_{i}u(t) + K_{i}(y(t) - \hat{y}_{u}(t)) + D_{i}\hat{d}(t)]$$

$$\hat{y}_{u}(t) = \sum_{i=1}^{r} u_{i}C_{i}\hat{x}_{u}(t)$$

$$\hat{d}(t) = \sum_{i=1}^{r} u_{i}G_{i}(y(t) - \hat{y}_{u}(t)) = \sum_{i=1}^{r} u_{i}G_{i}\widetilde{y}(t)$$
(12)

2.4. TS Fuzzy of Fuzzy Dedicated Observer (FDOS)

Since the system state variables cannot be directly measured, it is necessary to design a fuzzy observer to reconstruct the system state. Assuming that the state of the system model (1) is observable, and the following fuzzy observer Equation (13) is designed based on the T-S fuzzy model:

 R^i : If $z_1(t)$ is F_1^i and $z_2(t)$ is $F_2^i \dots$ and $z_{\psi}(t)$ is F_{ψ}^i , then

$$\hat{x}_{o}(t) = A_{i}\hat{x}_{o}(t) + B_{i}u(t) + N_{i}(y(t) - \hat{y}_{o}(t)) + D_{i}\hat{d}(t)
\hat{y}_{o}(t) = C_{i}\hat{x}_{o}(t) \qquad i = 1, 2, \dots, p$$
(13)

where $\hat{x}_o(t)$ is the estimated state by the fuzzy dedicated observer (FDOS), $\hat{y}_o(t)$ is the estimate output of the FDOS, $N_i \in K^{n \times 1}$ are the observer gain matrices. The inferred FDOS states Equation (14) can be obtained:

$$\dot{\hat{x}}_{o}(t) = \sum_{i=1}^{r} u_{i} [A_{i} \hat{x}_{o}(t) + B_{i} u(t) + N_{i} (y(t) - \hat{y}_{o}(t)) + D_{i} \hat{d}(t)]$$

$$\hat{y}_{o}(t) = \sum_{i=1}^{r} u_{i} C_{i} \hat{x}_{o}(t)$$
(14)

3. Design of Robust Scheduling Fault-Tolerant Controller

Assuming that the state of fuzzy system (1) is locally controllable, a local state feedback controller based on T-S fuzzy model (1) is designed for each subsystem based on the principle of PDC. The *j*th rule input by the controller is:

Rule *j*: If $g_1(t)$ is M_{1j} and $g_2(t)$ is M_{2j} ... and $g_k(t)$ is M_{kj} , then

$$u(t) = -L_j \hat{x}_o(t) - \overline{D}_j \hat{d}(t) + r(t) \qquad j = 1, 2, \dots, p$$
(15)

where $L_j \in \mathbb{R}^{m \times n}$ are the feedback gain vector of rule $j, r(t) \in \mathbb{R}^{m \times 1}$ is the reference input signals vector. Therefore, the overall Fuzzy Robust Scheduling Fault-Tolerant Controller (FRSFTC) can be expressed as:

$$u(t) = \sum_{j=1}^{p} u_j(g(t)) \Big[-L_j \hat{x}_o(t) - \overline{D}_j \hat{d}(t) + r(t) \Big] \qquad j = 1, 2, \dots, p$$
(16)

The preconditions of the designed fuzzy scheduling fault-tolerant controller are the same as those of the system model uncertainties. The *l*th rule is expressed by:

Rule *l*: If Δa_{11} is $N_{\Delta a_{11}}^l$ and ... and Δa_{nn} is $N_{\Delta a^{nn}}^l$ and Δb_{11} is $N_{\Delta b_{11}}^l$ and Δb_{nm} is $N_{\Delta b_{nm}}^l$, then

$$u(t) = \sum_{j=1}^{p} u_j(g(t)) \Big[-L_{jl} \hat{x}_o(t) - \overline{D}_j \hat{d}(t) + r(t) \Big]$$
(17)

 $u_i(g(t))$ is abbreviated as u_i , The control output of the proposed FRSFTC is procured as:

$$u(t) = \sum_{j=1}^{p} \sum_{l=1}^{s} u_{j} h_{l} \Big[-L_{jl} \hat{x}_{o}(t) - \overline{D}_{j} \hat{d}(t) + r(t) \Big]$$
(18)

Define the closed-loop equation of the system Equation (20) and estimation error Equation (19):

$$e_1(t) = x(t) - \hat{x}_0(t)$$
(19)

$$\dot{x}(t) = \sum_{i=1}^{r} \sum_{l=1}^{s} u_i h_l (A_i + \Delta \widetilde{A}_l) x(t) + \sum_{i=1}^{r} \sum_{l=1}^{s} u_i h_l D_i d(t) + \sum_{i=1}^{r} \sum_{l=1}^{s} u_i h_l (B_i + \Delta \widetilde{B}_l) u(t)$$
(20)

Substituting the Equation (18) into Equation (20), the closed-loop equation of the system Equation (20) in case of actuator failure can be represented by:

$$\dot{x}(t) = \sum_{i=1}^{r} \sum_{l=1}^{s} u_{i}h_{l}(A_{i} + \Delta \widetilde{A}_{l})x(t) + \sum_{i=1}^{r} \sum_{l=1}^{s} u_{i}h_{l}D_{i}d(t) + \sum_{i=1}^{r} \sum_{l=1}^{s} u_{i}h_{l}(B_{i} + \Delta \widetilde{B}_{l})\sum_{j=1}^{p} \sum_{l=1}^{s} u_{j}h_{l}\left[-L_{jl}\hat{x}_{o}(t) - \overline{D}_{j}\hat{d}(t) + r(t)\right]$$
(21)

$$Using \tilde{d}(t) = d(t) - \hat{d}(t)$$
(22)

According to Equations (19) and (22), it is easy to obtain that:

$$\dot{x}(t) = \sum_{i=1}^{r} \sum_{j=1}^{p} \sum_{l=1}^{s} u_{i}u_{j}h_{l} \Big[\Big((A_{i} + \Delta \widetilde{A}_{l}) - (B_{i} + \Delta \widetilde{B}_{l})L_{jl} \Big) x(t) + (B_{i} + \Delta \widetilde{B}_{l})L_{ji}e_{1}(t) + D_{j}\widetilde{d}(t) + (B_{i} + \Delta \widetilde{B}_{l})r(t) \Big]$$

$$(23)$$

Then the system state error is estimated as:

Then

$$\begin{aligned} \dot{e}_{2}(t) &= \dot{x}(t) - \dot{x}_{u}(t) \\ &= \sum_{i=1}^{r} \sum_{j=1}^{p} \sum_{l=1}^{s} u_{i} u_{j} h_{l} \Big[\Big((A_{i} + \Delta \widetilde{A}_{l}) - (B_{i} + \Delta \widetilde{B}_{l}) L_{jl} \Big) x(t) + (B_{i} + \Delta \widetilde{B}_{l}) e_{1}(t) \\ &+ D_{j} \widetilde{d}(t) + (B_{i} + \Delta \widetilde{B}_{l}) r(t) \Big] - \sum_{i=1}^{r} u_{i} [A_{i} \hat{x}_{u}(t) + B_{i} u(t) + K_{i} (y(t) - \hat{y}_{u}(t)) + D_{i} \hat{d}(t)] \\ &= \sum_{i=1}^{r} \sum_{j=1}^{p} \sum_{l=1}^{s} u_{i} u_{j} h_{l} \Big[\Big(\Delta \widetilde{A}_{l} - \Delta \widetilde{B}_{l} L_{jl} \Big) x(t) + \Big(A_{i} - K_{i} C_{j} + \Delta \widetilde{B}_{l} L_{jl} \Big) e_{2}(t) + \Delta \widetilde{B}_{l} r(t) + D_{j} \widetilde{d}(t) \Big] \end{aligned}$$
(26)

Assuming d(t) is time varying, we have:

$$\dot{\tilde{d}}(t) = \dot{d}(t) - \dot{\hat{d}}(t) = \dot{d}(t) - \sum_{i=1}^{r} u_i G_i C_i e_2(t)$$
(27)

Combining Equations (23), (24), (26) and (27), The following new augmented fuzzy system Equation (28) is obtained:

$$X(t) = \sum_{i=1}^{r} \sum_{j=1}^{p} \sum_{l=1}^{s} u_{i} u_{j} h_{l} (H_{ijl} + \Delta \widetilde{H}_{ijl}) x(t) + (S_{i} + \Delta \widetilde{S}_{l}) r(t) + Y \phi(t)$$
(28)

With: $X(t) = \begin{bmatrix} x(t) & e_1(t) & e_2(t) & \widetilde{d}(t) \end{bmatrix}$, $S_i = \begin{bmatrix} B_i & 0 & 0 & 0 \end{bmatrix}$, $\Delta \widetilde{S}_l = \begin{bmatrix} \Delta \widetilde{B} & \Delta \widetilde{B} & \Delta \widetilde{B} & 0 \end{bmatrix}$, $Y = \begin{bmatrix} 0 & 0 & 0 & I \end{bmatrix}$, $\phi(t) = \begin{bmatrix} \dot{d}(t) \end{bmatrix}$ and

$$H_{ijl} = \begin{bmatrix} A_i - B_i L_{jl} & B_i L_{jl} & 0 & D_j \\ 0 & A_i - N_i C_j & 0 & D_i \\ 0 & 0 & A_i - K_i C_j & D_i \\ 0 & 0 & -G_i C_i & 0 \end{bmatrix}, \\ \Delta \widetilde{H}_{ijl} = \begin{bmatrix} \Delta \widetilde{A}_l - \Delta \widetilde{B}_l L_{jl} & \Delta \widetilde{B}_l L_{jl} & 0 & 0 \\ \Delta \widetilde{A}_l - \Delta \widetilde{B}_l L_{jl} & \Delta \widetilde{B}_l L_{jl} & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$

Lemma 1. For uncertain parameters, the actuator fault fuzzy control system (28), if the inequality $\mu [TH_{ijl}T^{-1}] \leq -||T\Delta H_{ijl}T^{-1}||\max - \tau$ is true, Then, the system (28) is globally asymptotically stable, where τ is designed as positive value, T is an appropriate dimensional transformation symmetric matrix.

Proof. For system (28), it can be obtained according to Taylor formula

$$d\|TX(t)\|/dt \leq \sum_{i=1}^{r} \sum_{j=1}^{p} \sum_{l=1}^{s} u_{i}u_{j}h_{l}\left(\mu\left[TH_{ijl}T^{-1}\right] + \|T\Delta H_{ijl}T^{-1}\|\right)\|TX(t)\| \\ + \|\sum_{i=1}^{r} \sum_{l=1}^{s} u_{i}h_{l}T\left(S_{i} + \Delta \widetilde{S}_{l}\right)r(t)\| + \|\sum_{i=1}^{r} \sum_{l=1}^{s} u_{i}h_{l}TY\phi(t)\|$$

$$(29)$$

where
$$\mu \left[TH_{ijl}T^{-1} \right] = \lim_{\Delta t \to 0^+} \frac{\|I + TH_{ijl}T^{-1}\Delta t\| - 1}{\Delta t} = \lambda_{\max} \left(\frac{TH_{ijl}T^{-1} + \left(TH_{ijl}T^{-1} \right)^*}{2} \right)$$
 (30)

where $\lambda_{\max}(.)$ denotes the largest eigen value, * represents the conjugate transpose. Assuming the fault is bounded, let $||d(t)|| \le d_{a\max}$, $0 \le d_{a\max} < +\infty$. since $\phi(t) = d(t)$. accordingly, we can get that:

$$\|\phi(t)\| \le d_{a\max}, 0 \le d_{a\max} < +\infty \tag{31}$$

If the following inequality

$$\mu \left[TH_{ijl}T^{-1} \right] \le -\|T\Delta H_{ijl}T^{-1}\|\max - \tau$$
(32)

are satisfied, where τ is a nonzero positive constant. According to Equations (29) and (31), it can be obtained that:

$$\frac{d\|TX(t)\|}{dt}\exp(\tau(t-t_0)) \le \sum_{i=1}^{p} \sum_{l=1}^{s} u_i h_l(\|T(S_i + \Delta \widetilde{S}_l)r(t)\| + \|TY\phi(t)\|)\exp(\tau(t-t_0))$$
(33)

where $t_0 < t$ is an arbitrary initial time. If (31) is satisfied, then the system (28) is globally asymptotically stable when $t \to \infty$, $||X(t)|| \to 0$. supposing r(t) = 0, $\phi(t) = 0$ and $r(t) \neq 0$, $\phi(t) \neq 0$, According to (31) and (32), it is easy to see that:

$$\|T\widehat{Y}\phi(t)\| \ge \max_{i} \|T\widehat{Y}\phi(t)\|_{\max} \ge \|TY\phi(t)\|$$
(34)

$$\|TX(t)\| \le \|TX(t_0)\|e^{-\tau(t-t_0)} + \frac{\|T(\widehat{S}_i + \Delta \widetilde{S}_l)r(t)\|}{\tau}(1 - e^{-\tau(t-t_0)}) + \frac{\|T\widehat{Y}\phi(t)\|}{\tau}(1 - e^{-\tau(t-t_0)})$$
(35)

where $||T(\widetilde{S}_i + \Delta \widetilde{S}_l)r(t)|| \ge \max_i ||T(S_i + \Delta \widetilde{S}_l)r(t)||_{\max} \ge ||T(S_i + \Delta \widetilde{S}_l)r(t)||$. From (35) is bounded, when r(t) and $\phi(t)$ are bounded, we can get that the system is also bounded, therefore the system is stable.

Theorem 1. For the fuzzy control system as presented by system (28), supposing that there are matrices X_i , M_{a11} , W_j and O_i such that the controller and observer gains of the fuzzy system are set to $L_j = W_{jl}M_{a11}^{-1}$, $N_i = P_{a22}^{-1}O_i$ and $\overline{E}_i = P_2^{-1}X_i$, and satisfy the following inequality:

$$M_{a11}A_{i}^{T} + A_{i}M_{a11} - (B_{i}W_{jl})^{T} - B_{i}W_{jl} < 0$$

$$A_{i}^{T}P_{a22} + P_{a22}A_{i} - (O_{i}C_{j})^{T} - O_{i}C_{j} < 0$$

$$H_{bi}^{T}P_{2} + P_{2}H_{bi} - (X_{i}\overline{C}_{j}) - X_{i}\overline{C}_{j} < 0$$
(36)

Then the closed-loop system (28) is globally asymptotically stable.

Proof. Fault-tolerant control is designed to find controller and observer gain L_j , N_i , K_i and G_i in order that the asymptotic convergence of X(t) tends to zero, if r(t) = 0, $\phi(t) = 0$ and to ensure a bounded state based on (31), if $r(t) \neq 0$, $\phi(t) \neq 0$, this problem is transformed into finding matrix P verifying V(t) < 0.

Define the Lyapunov function as follows:

$$V(t) = X(t)^{T} P X(t)$$
(37)

$$PH_{ijl} + H_{ijl}^T P < 0 \forall i, j, l$$
(38)

From system (28), matrix H_{ijl} , ΔH_{ijl} , S_i , ΔS_l , P and Y can be expressed as:

$$\begin{split} H_{ijl} &= \begin{bmatrix} H_{aijl} & H_{cij} \\ 0_{2\times2} & H_{bi} - \overline{E}_i \overline{C}_j \end{bmatrix}, \Delta H_{ijl} = \begin{bmatrix} \Delta \overline{A}_{ila} & 0_{2\times2} \\ \Delta \overline{A}_{ilb} & 0_{2\times2} \end{bmatrix}, S_i = \begin{bmatrix} \overline{B}_i \\ 0 \end{bmatrix}, \Delta \widetilde{S}_l = \begin{bmatrix} \Psi_a \\ \Psi_b \end{bmatrix}, \Psi_a = \begin{bmatrix} \Delta \widetilde{B}_l \\ \Delta \widetilde{B} \end{bmatrix}, \\ \Psi_b &= \begin{bmatrix} \Delta \widetilde{B} \\ 0 \end{bmatrix}, H_{aijl} = \begin{bmatrix} A_i - B_i L_{jl} & B_i L_{jl} \\ 0 & A_i - N_i C_j \end{bmatrix}, H_{cij} = \begin{bmatrix} 0 & D_j \\ 0 & D_j \end{bmatrix}, H_{bi} = \begin{bmatrix} A_i & D_i \\ 0 & 0 \end{bmatrix}, \overline{E}_i = \begin{bmatrix} K_i \\ G_i \end{bmatrix}, \\ \overline{C}_j &= \begin{bmatrix} C_j & 0 \end{bmatrix}, P = \begin{bmatrix} P_1 & 0_{2\times2} \\ 0_{2\times2} & P_2 \end{bmatrix}, Y = \begin{bmatrix} 0 \\ \overline{Y} \end{bmatrix}, \overline{Y} = \begin{bmatrix} 0 \\ I \end{bmatrix}, \Delta \overline{A}_{ijla} = \begin{bmatrix} \Delta \widetilde{A}_l - \Delta \widetilde{B}_l L_{jl} & \Delta \widetilde{B}_l L_{jl} \\ \Delta \widetilde{A}_l - \Delta \widetilde{B}_l L_{jl} & \Delta \widetilde{B}_l L_{jl} \end{bmatrix}, \\ \Delta \overline{A}_{ijlb} &= \begin{bmatrix} \Delta \widetilde{A}_l - \Delta \widetilde{B}_l L_{jl} & \Delta \widetilde{B}_l L_{jl} \\ 0 & 0 \end{bmatrix} \end{split}$$

Hence, matrix (37) can be re-expressed as:

$$P_1 H_{aijl} + H_{aijl}^T P_1 < 0 \forall i, j$$

$$\tag{39}$$

$$P_2(H_{bi} - \overline{E}_i \overline{C}_j) + (H_{bi} - \overline{E}_i \overline{C}_j) P_2 < 0 \forall i, j$$

$$\tag{40}$$

Equations (39) and (40) are a set of Nonlinear Matrix Inequalities; it is not a linear Matrix Inequality, assuming $P_1 = diag(P_{a11}, P_{a22})$, applying the change of variables $W_j = M_{a11}L_j$, $O_i = P_{a22}N_i$ and $X_i = P_2\overline{E}_i$, multiplying (39) on the left by $M_{a11} = P_{a11}^{-1}$ and right by $M_{a11} = P_{a11}^{-1}$, the LMIS (36) in the theorem are acquired. \Box

4. Application Examples of WECS

4.1. Dynamic Mathematical Model of WECS

According to Betz theory, the mechanical power captured by the wind turbine from the wind energy is:

$$P_{wt} = 0.5\rho R^2 V^3 C_p(\lambda,\beta) \tag{41}$$

where ρ represents the air density, *R* denotes the rotor-plane radius, *V* is the wind speed, β is the pitch angle, λ is the tip speed ratio (TSR), Ω_r is the turbine rotational speed of the low-speed shaft, C_p is the power coefficient to convert wind energy into mechanical energy. TSR λ is the ratio of blade tip linear velocity to wind speed of a wind turbine, which is defined by $\lambda = \Omega_r R/V$. The output torque of the wind turbine T_{wt} can be expressed as:

$$T_{wt} = P_{wt} / \Omega_r = 0.5\rho R^2 V^3 C_p(\lambda,\beta) / \Omega_r$$
(42)

When the wind speed is a constant, the mechanical power output of the wind turbine only depends on the power coefficient C_p . If the pitch Angle β stays the same, the power coefficient C_p is only determined by the TSR λ . For a particular wind turbine, there is a single optimal TSR λ_{opt} . At this time, C_{pmax} is defined as the maximum wind energy capture coefficient. The maximum capture of wind energy can be achieved by adjusting the electromagnetic torque of the generator to follow the change of wind speed to make it reach the maximum. Fixed pitch control is adopted under rated wind speed that is $\beta = 0$ When the TSR $C_p(\lambda, \beta) = C_p(\lambda)$, $C_{pmax} \approx 0.48$, namely the optimal TSR.

WECS is mainly composed of the wind turbine, transmission system, generator, and power grid. The wind turbine captures wind energy, and converts it into mechanical energy via the wind turbine rotation, which drives the rotor of the doubly fed induction motor to rotate by the transmission system, thus generating electricity, through the converter, to the grid. The overall block diagram of wind energy conversion system with a DFIG is depicted in Figure 1.



Figure 1. Overall block diagram of the wind energy conversion system (WECS) with a doubly fed induction generators (DFIG).

Using the kinematics equation of the transmission system [20–22], The dynamics model of the wind power generation system Equation (43) can be obtained as follows:

$$\frac{d\Omega_r}{dt} = \left(\frac{D_r}{J_r} + \frac{K_{opt}}{J_r}\Omega_r\right)\Omega_r - \frac{n_b}{J_r}T_h$$

$$\frac{d\Omega_g}{dt} = -\frac{D_g}{J_g}\Omega_g + \frac{1}{J_g}T_h - \frac{1}{J_g}T_g$$

$$\frac{dT_h}{dt} = \frac{1}{n_b}\left(K_{ls} - \frac{D_rD_{ls}}{J_r} + \frac{D_{lse}K_{opt}}{J_r}\Omega_r\right)\Omega_r - \frac{1}{n_b^2}\left(K_{ls} - \frac{D_gD_{ls}}{J_g}\right)\Omega_g - D_{ls}\left(\frac{1}{J_r} + \frac{1}{n_b^2J_g}\right)T_h + \frac{D_{ls}}{n_b^2J_g}T_g$$

$$\frac{dT_g}{dt} = -\frac{1}{\tau_g}T_g + \frac{1}{\tau_g}T_{g,ref}$$
(43)

where $K_{opt} = 0.5\rho\pi R^5 C_{pmax} / \lambda_{opt}^2$, D_r and D_g represent the damping constants of the rotor and the generator respectively, τ_g is the time constant, K_{ls} denotes the equivalent torsional stiffness of the low-speed shaft, D_{ls} is the damping constants of the equivalent low-speed shaft, T_h is the high-speed shaft torque, J_r is the rotor moment of inertia, J_g is the generator moment of inertia, T_g is the generator speed and n_b is the generator torque, $T_{g,ref}$ is the required generator torque, Ω_g is the mechanical generator speed and n_b is the generator ratio.

From the dynamics model (43), the standard form of WECS equation of state can be expressed as:

$$\dot{x}(t) = A(x)x(t) + Bu(t)$$

$$y(t) = C(x)x(t)$$
(44)

with $x(t) = \begin{bmatrix} x_1 & x_2 & x_3 & x_4 \end{bmatrix}^T = \begin{bmatrix} \Omega_r & \Omega_g & T_h & T_g \end{bmatrix}^T$, $u(t) = T_{g,ref}$,

$$y(t) = \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} \Omega_r \\ \Omega_g \end{bmatrix}, a_1 = \frac{1}{n_b} (K_{ls} - \frac{D_r D_{ls}}{J_r}), a_2 = -\frac{1}{n_b^2} (K_{ls} - \frac{D_g D_{ls}}{J_g}), a_3 = -D_{ls} (\frac{1}{J_r} + \frac{1}{n_b^2 J_g})$$

4.2. T-S Fuzzy Description of WECS

T-S fuzzy is applied to WECS. By looking at the function of the system matrix A(x), the premise variable $z_1(t) = \Omega_r$ and $z_2(t) = \Omega_g$ are defined. Then, the membership functions of $z_1(t)$ and $z_2(t)$ are selected $A(x) = \sum_{i=1}^r u_i(z(t))A_i$, $B(x) = \sum_{i=1}^r u_i(z(t))B_i$, $C(x) = \sum_{i=1}^r u_i(z(t))C_i$.

To simplify, the membership function of the two fuzzy subsets can be expressed by Equation (45):

$$F_j(z_j(t)) = \frac{-z_{j\min}}{z_{j\max} - z_{j\min}} \left(\frac{1}{z_{j\max} - z_{j\min}}\right) z_{jt}$$

$$\overline{F}_j(z_j(t)) = 1 - F_j(z_j(t)) (j = 1, 2)$$
(45)

where the variable z_{jt} is bounded by its upper value z_{jmin} and lower value z_{jmax} . The membership functions of $z_1(t)$ are depicted in Figure 2, and each membership function also indicates the model uncertainty of each subsystem. The membership functions of $z_2(t)$ are implemented in the same way.



Figure 2. The membership functions of $z_1(t)$.

The T-S fuzzy model with uncertain parameters and actuator failure of WECS (44) can be expressed by the following 4 rules:

Rule 1: If $z_1(t)$ is F_1 and $z_2(t)$ is F_2

Then
$$\dot{x}(t) = (A_1 + \Delta A_1)x(t) + (B_1 + \Delta B_1)u(t) + D_1d(t)$$
 (46)

Rule 2: If $z_1(t)$ is F_1 and $z_2(t)$ is \overline{F}_2

Then
$$\dot{x}(t) = (A_2 + \Delta A_2)x(t) + (B_2 + \Delta B_2)u(t) + D_2d(t)$$
 (47)

Rule 3: If $z_1(t)$ is \overline{F}_1 and $z_2(t)$ is F_2

Then
$$\dot{x}(t) = (A_3 + \Delta A_3)x(t) + (B_3 + \Delta B_3)u(t) + D_3d(t)$$
 (48)

Rule 4: If $z_1(t)$ is \overline{F}_1 and $z_2(t)$ is \overline{F}_2

Then
$$\dot{x}(t) = (A_4 + \Delta A_4)x(t) + (B_4 + \Delta B_4)u(t) + D_4d(t)$$
 (49)

Therefore, the global fuzzy model of wind energy conversion system is presented by:

$$\dot{x}(t) = \sum_{i=1}^{4} u_i(z(t)) \Big[(A_i + \Delta A_i) x(t) + (B_i + \Delta B_i) u(t) + D_i d(t) \Big]$$

$$y(t) = \sum_{i=1}^{4} u_i(z(t)) C_i x(t) = 1, 2, 3, 4$$
(50)

where

 ΔA_i and ΔB_i represent the system parameter bounded uncertainties, with $\Delta B_i = 0$. The change of the parameter uncertainties ΔJ_g within 30% and 50% of nominal value is considered. From Equation (6), we can get c = 6, s = 64. Therefore, the fuzzy uncertainty is given by $\Delta A_i = \sum_{l=1}^{64} h_l \Delta \widetilde{A}_l$. Then, Equation (28) of WECS fuzzy model can be presented as:

$$\dot{x}(t) = \sum_{i=1}^{4} \sum_{l=1}^{64} u_i h_l \Big[(A_i + \Delta \widetilde{A}_l) x(t) + B_i u(t) + D_i d(t) \Big]$$
(51)

From (18), The inferred output of the FRSFTC is:

$$u(t) = \sum_{j=1}^{4} \sum_{l=1}^{64} u_j h_l \Big[-L_{jl} \hat{x}_o(t) - \overline{D}_j \hat{d}(t) + r(t) \Big]$$
(52)

5. Simulation Analysis

Based on system model (50), a WECS with low power (6KW) and high speed is simulated in MATLAB/Simulink environment (R2014a, MathWorks, Natick, MA, USA). The parameters of the simulation system are depicted in Table 1:

Parameter Names	Values
Rated power P_n	6 KW
Rated voltage V_s	220 V
Rated speed w_s	$100 \pi rad/s$
Air density ρ	1.25 kg/m^3
Blade length R	2.5 m
Transmission efficiency η	0.95
Pole pairs <i>p</i>	2
Rated electromagnetic torque T_g	40 N · m
Inertia of the generator J_g	0.0092 kg \cdot m ²
Inertia of the rotor J_r	$3.6 \text{ kg} \cdot \text{m}^2$

 Table 1. Simulation Parameters.

As the actuator of the entire WECS, the transmission system is mainly caused by faults such as deviation, drift and so on [23,24]. To facilitate the research and simulation, this paper only considers the actuator faults and does not involve unknown external interference. Accordingly, the drift, deviation and mixed fault of the actuator are considered in the simulation design, and as shown below:

$$d(t) = \begin{cases} 4\sin \pi t & 50 \text{ s} \le t < 60 \text{ s} \\ 3 & 60 \text{ s} \le t < 70 \text{ s} \\ 2.5 + 2\sin 0.5\pi t & 70 \text{ s} \le t < 80 \text{ s} \end{cases}$$
(53)

Case 1: The change of the parameter uncertainties ΔJ_g within 30% of nominal value is considered.

The controller is tested according to the random change of wind speed. Figure 3 is the wind speed waveform, Figure 4 is the actual and estimated value of time-varying fault signal, it can be seen that the designed observer can rapidly and accurately reconstruct the fault information of the system. When the system fails, the operation of the system with and without FRSFTC strategy is shown in Figures 5–8, Figure 9 shows the power coefficient when FRSFTC is adopted in the case of actuator faults.

Comparing (a) and (b) in Figures 5 and 6, it can be seen that between t = 50 s and t = 80 s, when the actuator of the system fails, both the trajectories of Ω_r and Ω_g have abrupt changes, and the oscillation amplitude increases and cannot be maintained in the optimal position. However, under the FRSFTC strategy, the system of low- and high-speed fluctuation range is reduced. The system failure of gear and bearing fault impact and oscillation are greatly reduced.



Figure 4. Actuator faults signal d(t) and its estimate.



Figure 5. The trajectories of Ω_r . (a) Without using the FRSFTC strategy; (b) Using the FRSFTC strategy.



Figure 6. The trajectories of Ω_g . (a) Without using the FRSFTC strategy; (b) Using the FRSFTC strategy.

Case 2: The change of the parameter uncertainties ΔJ_g within 50% of nominal value is considered. According to simulation results depicted in Figures 7 and 8, it can be seen from the comparison between (a) and (b) when the parameter uncertainty ΔJ_g is within 50% of nominal value, the designed

FRSFTC strategy can still reduce the fluctuation range of the system at low speed and high speed and achieve a good fault-tolerance control effect.

The power coefficient is shown in Figure 9, and the $C_{pmax} \approx 0.48$, which shows that when the actuator fails, the WECS can achieve maximum wind energy capture.

In summary, the simulation results show that when the actuator fault occurs, considering the uncertainty of the system, the proposed FRSFTC strategy can achieve the maximum wind energy capture at the rated wind speed and improve the use rate of the wind turbine when the actuator fault occurs.



Figure 7. The trajectories of Ω_r . (a) Without using the FRSFTC strategy; (b) Using the FRSFTC strategy.



Figure 8. The trajectories of Ω_g . (a) Without using the FRSFTC strategy; (b) Using the FRSFTC strategy.



Figure 9. Power coefficient.

6. Conclusions

This paper presents the idea of robust fault-tolerant control based on system state estimation and fault reconstruction for parameter-uncertain nonlinear systems with actuator faults. We first introduce the T-S fuzzy model of the nonlinear system, and then a fuzzy scheduling fault-tolerant controller based on observer fault reconstruction is proposed for the uncertainty, unmeasurable state, and actuator fault of the system. The gain of controller and observer is obtained by solving LMIS. Finally, according to the Taylor series and Lyapunov stability theory, the sufficient and necessary conditions for the stability of the closed-loop system in the event of actuator failure are given, and fault-tolerance integrity of the system is realized. Finally, the WECS is taken as an example for analysis. Simulation results show that considering the uncertainty of the system, when the actuator fails, the FRSFTC can improve the efficiency of wind energy capture at a rated wind speed while ensuring the normal operation of all states of the system. The feasibility and effectiveness of the proposed fault-tolerant control strategy are verified.

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